

**LECTURE NOTES ON  
ENERGY CONVERSION-I**

**SEMESTER- 4<sup>th</sup>**

**PREPARED BY  
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# SINGLE PHASE TRANSFORMER

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Transformer is a static piece of device by means of which electric power <sup>or energy</sup> in one circuit is transformed into electric power of same frequency in another circuit, <sup>by mutual induction</sup> It can raise or lower the voltage in circuit but with a corresponding decrease or increase in current.

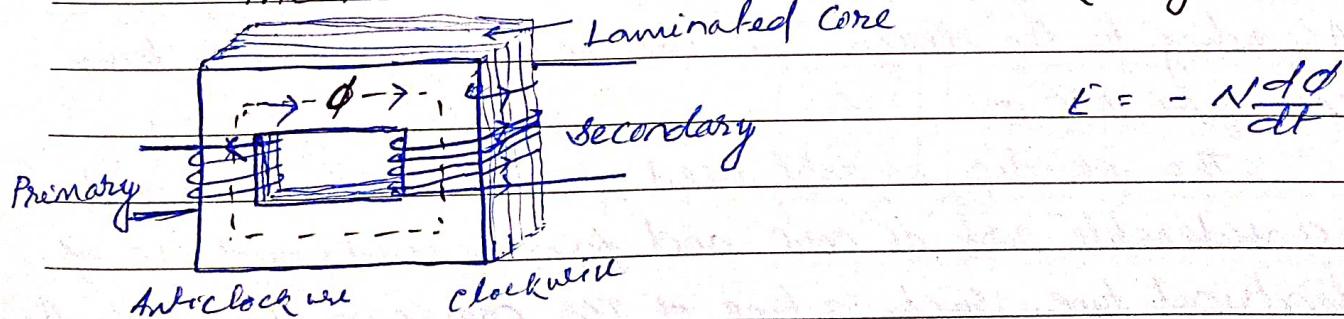
## Working Principle:

It works on the principle of mutual induction of two coils or Faraday law's of Electromagnetic induction

If one coil of transformer is connected to a source of an alternating voltage, an alternating flux is setup in the laminated core, most of which is linked with the other coil in which it produces mutually induced emf.

In brief Transformer is a device that

- \* Transforms electric power from one circuit to another.
- \* It does so without a change of frequency.
- \* Transfer off with the principle of Electromagnetic Induction.
- \* The two electrical circuit are linked by mutual induction.



## Construction:

The simple elements of a transformer consist of two coils having mutual inductance of laminated



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steel core. The two coils are insulated from each other and from steel core. Device also needs some suitable container for the assembled core and windings, a medium with which the core and windings can be insulated. Suitable bushings either porcelain, oil-filled or capacitor-type for insulating and bringing out the terminals of windings from the tank. The T/F core is made up of or constructed of transformer sheet steel laminations assembled to provide a continuous magnetic path with a minimum of air-gap included.

✓ The steel should have high permeability and low hysteresis loss. For thin to happen, the steel should be made up of high silicon content and must also be heat treated. By effectively laminating the core, the eddy current losses can be reduced, the laminations being insulated from each other by a light coat of core-plate varnish or by an oxide layer on the surface.

### Type of Transformer:

According to the design T/F can be classified into two:

#### ④ Core type T/F :

✓ The winding or coil used for this T/F surround a considerable part of core and form around and are of cylindrical type. Such a type of T/F can be applicable for small sized and large sized T/Fs.

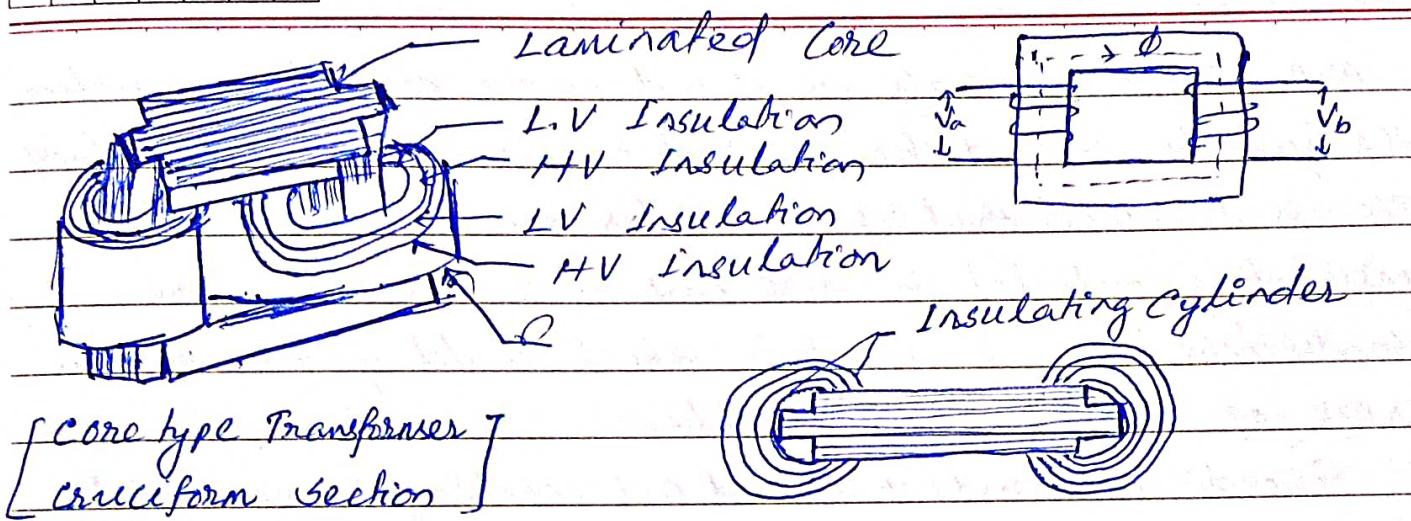
In small sized type the core will be rectangular, and coil used are cylindrical, which are either circular or rectangular in form.



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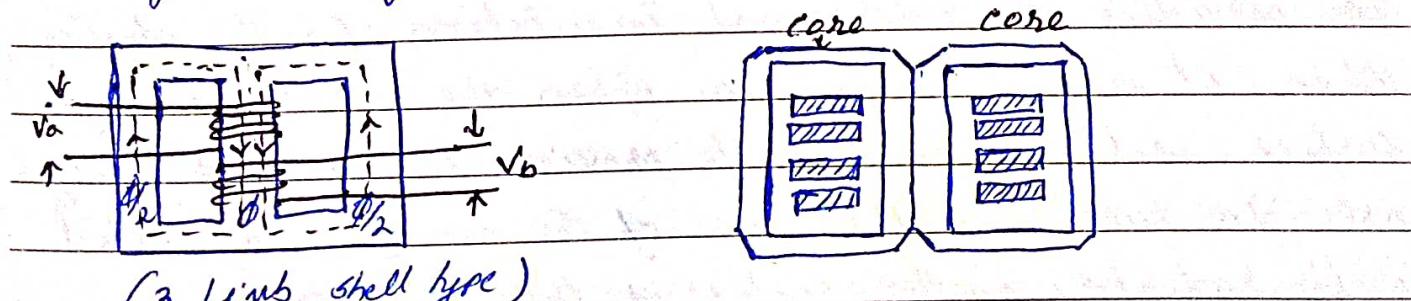


→ The cylindrical coils will have different layers and each layer will be insulated from each other with the help of material like paper, cloth, mica board and so on.

\* These are popular in High Voltage applications like Distribution<sup>F</sup>, Paper T/F and Auto T/F

### (B) Shell-Type T/F :

In shell-type T/F, the core surrounds a considerable portion of the windings. The coils are form-wound but are multilayer disc type usually wound in the form of pancakes.



Paper is used to insulate the different layers of the multi-layer discs. The whole winding consist of discs stacked with insulation spaces between the coils. These form horizontal cooling & insulating ducts.



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Both H.V & L.V coils are wound on the central limb surface. The quantity of conductor required for windings of shell type T/F is less than that of a similar core type T/F. The insulating material is more here as the windings are sandwiched on each other which could give rise to a risk of short-circuit faults.

Forced air and/or forced oil cooling is essential in shell type T/F as heat generated during working can't get dissipated easily from windings due to surrounding yoke and limbs.

Shell type T/F are mainly used for Low Voltage (LV) applications and are very often used in low voltage power circuits as well as in electronics circuits.

→ The sides of insulating tank, oil used in the T/F provide needed to circulate through the device and cool the coils. It is also responsible for providing the additional insulation for the device when it is left in the air. When the smooth tank surface will not be able to provide the needed cooling air. In such cases the sides of the tank are assembled with radiators on the sides of the device.

→ Breather is used to absorb the moisture present in air before entering into the tank during expansion and contraction oil in the tank.



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Type according to the type of cooling employed ~

(A) Oil filled self cooled type ~

- It is used by small and medium sized distribution T/F. The assembled windings and core of such T/F are mounted in a welded, oil-tight steel tanks provided with a steel cover. The tank is filled with purified, high quality insulating oil as soon as the core is put back at its proper place.

The oil helps in transforming the heat from core and the winding to the surroundings. For smaller size tanks are, usually, smooth surfaced but for large size T/F greater surface area is needed and is achieved by frequently corrugating the case.

(B) Oil filled water cooled type ~

It is same as oil filled self cooled type, only difference is that a cooling coil is mounted near the surface of the oil, through which cold water keep circulating.

This design is usually implemented on T/F's that are used in high voltage transmission lines.

(C) Air Blast type ~

This type is used for T/F's that use voltage below 25 KV. The transformer is housed in a thin sheet metal box open at both ends through which air is blown from the bottom to the top.

Sine of Angle = Cosine of its complement



$$\sin \theta = \cos(90^\circ - \theta)$$

$$\sin \theta = \sin(90^\circ - \theta) \text{ or } \sin(90^\circ + \phi)$$

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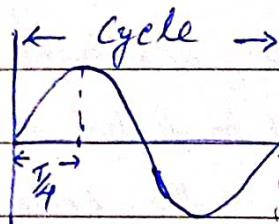
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## E.M.F Equation of a Transformer

Let  $N_1$  = No of turns in primary  $N_2$  = No of turns in Secondary

$\Phi_m$  = Max flux in core in Weber,  $B_m = B_m \times A$

$f$  = Frequency of a.c input in Hz.



flux increases from zero to max

value  $\Phi_m$  in one quarter of the cycle

$$\therefore \text{Average rate of change of flux} = \frac{\Phi_m}{T/4} \quad T = \frac{1}{f}$$
$$= 4f\Phi_m \text{ Wh/s or Volt}$$

The rate of change of flux per turns means induced emf in volt. i.e

$$\text{Average emf/turn} = 4f\Phi_m \text{ Volt.}$$

If  $\Phi$  varies sinusoidally, then rms value of induced emf is obtained by multiplying the average value by form factor

$$\text{RMS Value} = \text{Form factor} \times \text{Average value}$$

$$\text{RMS Value of emf/turn} = 1.11 \times 4f\Phi_m = 4.44f\Phi_m \text{ Volt}$$

$$\text{RMS Value of the induced emf in the whole of primary winding} \\ = \text{Induced emf/turn} \times \text{No of primary turns}$$

$$E_1 = 4.44f N_1 \Phi_m = 4.44f N_1 B_m A \quad \text{--- (1)}$$

Similarly rms value of the emf induced in secondary is

$$E_2 = 4.44f N_2 B_m A \quad \text{--- (2)}$$

From Eqn (1) and (2)

$$\left| \frac{E_1}{N_1} = \frac{E_2}{N_2} = 4.44f B_m A \right|$$

OR  $\theta = \theta_m \sin \omega t$   $e = -N \frac{d\theta}{dt} = -N \dot{\theta}_m (\theta_m \sin \omega t) = -N \theta_m \cos \omega t \cdot \omega = -N \theta_m \omega \sin(90^\circ - \omega t)$



$\Rightarrow e = N \theta_m \omega \sin(\omega t - 90^\circ)$

RMS value  $E_r = \frac{N \theta_m \omega}{\sqrt{2}} = \frac{N \theta_m (2\pi f)}{\sqrt{2}}$  Date / /

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\* Voltage Transformation Ratio ( $K$ ) =  $\frac{E_2}{E_1} = \frac{N_2}{N_1}$

\* Turn Ratio =  $\frac{N_1}{N_2}$

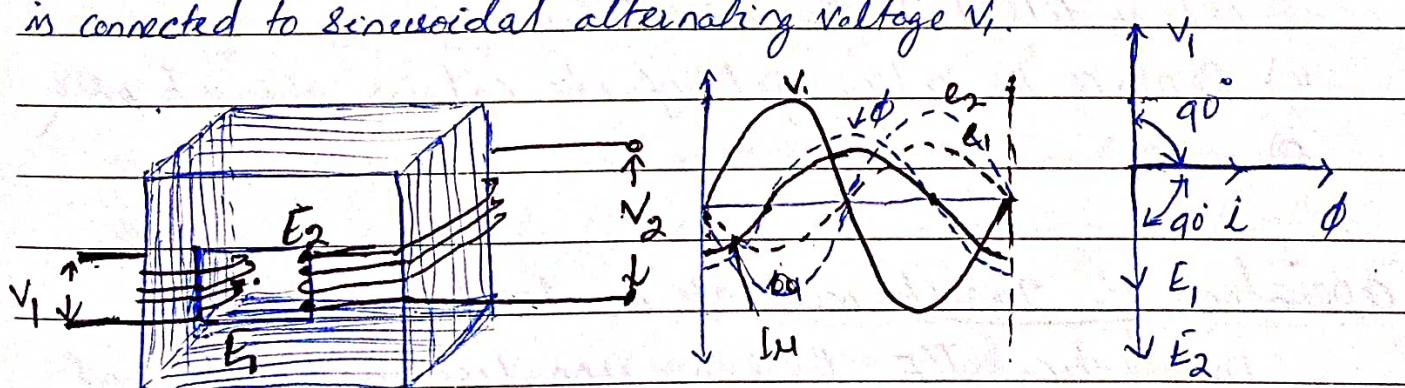
For ideal transformer input VA = output VA

$$V_1 I_1 = V_2 I_2 \text{ or } \frac{I_2}{I_1} = \frac{V_1}{V_2} = \frac{1}{K} \quad \left. \begin{array}{l} \text{Here } E_1 = V_1 \\ E_2 = V_2 \end{array} \right\}$$

### Ideal Transformer :-

An ideal T/F is one which has no losses i.e. its winding have no ohmic resistance, there is no magnetic leakage and hence which has no  $I^2 R$  and core losses. In other word, an ideal T/F consist of two purely inductive coils wound on a loss-free core.

consider an ideal T/F whose secondary is open and primary is connected to sinusoidal alternating voltage  $V_1$ .



The potential difference causes an alternating current to flow in primary.

Since primary coil is purely inductive and there is no output (secondary being open) the primary draws the magnetising current ( $I_M$ ) only.



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$I_m$  is small in size and lags  $V_1$  by  $90^\circ$ .  $I_m$  produces an alternating flux  $\Phi$  which is proportional to the current and is in phase with it. Because of change in flux with primary and secondary windings, it produces self-induced emf ( $E_1$ ) equal and opposite to  $V_1$ . Similarly in secondary side emf ( $E_2$ ) is induced, known as mutually induced emf is antiphase with  $V_1$ .

### Care and Maintenance of Transformer :-

- 1) Oil level in the oil cap must be checked on a monthly basis.
- 2) Keep the breathing holes in the silica gel breather clean to ensure proper breathing action.
- 3) If T/F has oil filling bushing, make sure that the oil is filled to the correct level.
- 4) Replace the silica gel if its colour changes to pink.
- 5)

### Operation of Transformer at no-load.

- In practical T/F there is iron loss in the core and copper loss in the winding and these are not entirely negligible.
- When T/F is on no-load, the primary input current is not wholly reactive. Primary current ( $I_{1n}$ ) has to supply no load
- 1) Iron losses in the core i.e. hysteresis loss ( $P_{1n}$ ) and Eddy current losses ( $P_{2n}$ ) and a very small



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amount of copper loss in primary. So  $I_o$  is not at  $90^\circ$  behind  $V_1$ , but lags it by angle  $\phi_o < 90^\circ$

$$\text{No Load Power Input } (P_{no}) = V_1 I_o \cos \phi_o$$

$I_o$  has two components: one in phase

with  $V_1$ , is known as active or working

or iron loss components  $I_{oI}$ .

$$|I_{oI}| = I_o \cos \phi_o$$

\*  $I_{oI}$  supplies the iron loss and small quantity of primary cu. loss.

ii) Other component is quadrature with  $V_1$  and is known as magnetising component  $I_{oM}$ , because its work is to sustain the alternating flux in core,

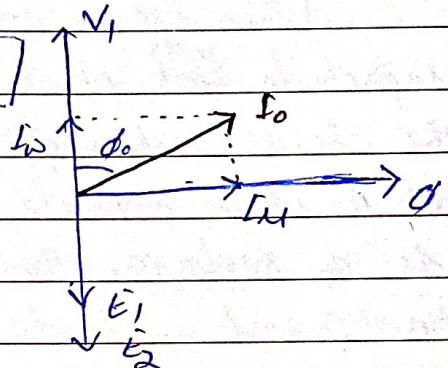
$$|I_{oM}| = I_o \sin \phi_o$$

$$\text{Here } I_o = I_{oI} + I_{oM}, \text{ hence } I_o = \sqrt{I_{oM}^2 + I_{oI}^2}$$

\*  $I_o$  is very small and is  $1\%$  of full load current.

\* As  $I_o$  is very small, so no-load primary cu. loss is negligible small hence No-load Primary Input is practically equal to the iron loss in T/F.

\* Core loss or iron loss is responsible for shifting the current vector, hence  $\phi_o$  is known as hysteresis angle of advanced



[phasor diagram of  
no-load T/F]



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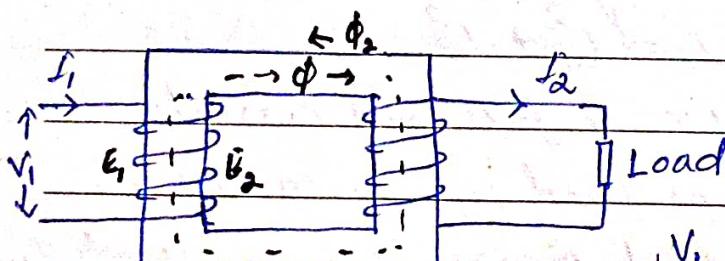
## Transformer on Load

When the secondary is loaded, the current  $I_2$  is set up. Magnitude <sup>and</sup> phase of  $I_2$  with respect to  $V_2$  is determined by the characteristics of Load.  $I_2$  sets up its own emf ( $E_2'$ ) and (A) it is opposite to the main flux.

As  $\phi_2$  reduces the main flux ( $\Phi$ ) momentarily, hence primary back emf  $E_1$  tends to be reduced. For a moment  $V_1$  gain the upper hand and primary current increases and this additional current be  $I_2'$  and is antiphase with  $I_2$  and sets up its own flux  $\phi_2'$  opposite to  $\phi_2$ .

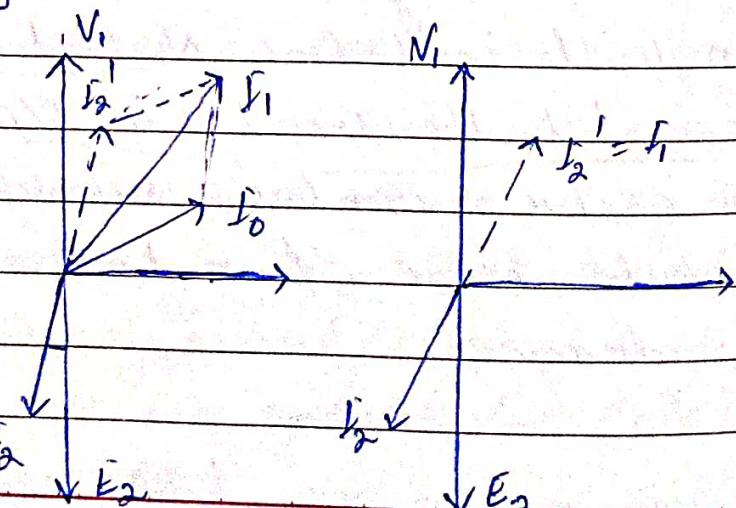
\* Net flux passing through the core is approximately the same as at no-load. Hence core loss is also practically same under all load condition.

$$\text{As } \phi = \phi' \therefore N_2 I_2 = N_1 I_2' \therefore I_2' = \frac{N_2}{N_1} I_2 = K I_2$$



$$I_1 = I_0 + I_2$$

$$I_2' = K I_2$$



(a) Resistive Load ]

(b) - Lagging Load ] [ (c) - if  $I_0$  is neglected ]



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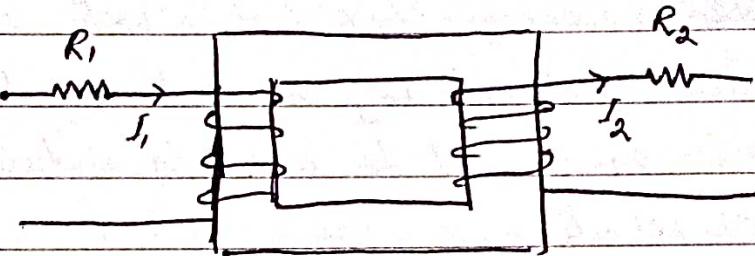
$\phi_1$  is slightly greater than  $\phi_2$ . If we neglect  $I_o$  as compared to  $I_2'$  then  $\phi_1 = \phi_2$

$$\text{Hence } N_1 I_2' = N_2 I_2 \Rightarrow N_1 I_2 \quad \therefore \frac{I_2'}{I_2} = \frac{N_2}{N_1} = \frac{r_1}{r_2} = k$$

(As  $I_1 = I_2'$ )

## Equivalent Resistance, Reactance and Impedance

Primary & secondary winding, having resistance  $R_1$ ,  $R_2$  respectively is shown in figure.



The resistance of the two winding can be transferred to any one of the two winding. It makes calculations very simple and easy because one has to work one winding only.

where  $R'_1$  = The equivalent primary resistance as referred to secondary  
 $\& R'_2$  = " " secondary " " " primary.

These can be calculated by considering the same losses ( $I^2 R$  loss)

$$I_1^2 R'_2 = I_2^2 R_2 \Rightarrow R'_2 = \left(\frac{I_2}{I_1}\right)^2 R_2 \Rightarrow \boxed{R'_2 = \frac{R_2}{k^2}}$$

Similarly  $\boxed{R'_1 = k^2 R_1}$

$\therefore$  Effective or equivalent resistance of T/F as referred to primary  $(R_{eq}) = R_1 + R'_1$   
 $= R_1 + R_2/k^2$

$$\begin{aligned} \text{R. } & " \quad " \quad \text{to secondary} (R_{eq}) = R_2 + R'_2 \\ & = R_2 + k^2 R_1 \end{aligned}$$

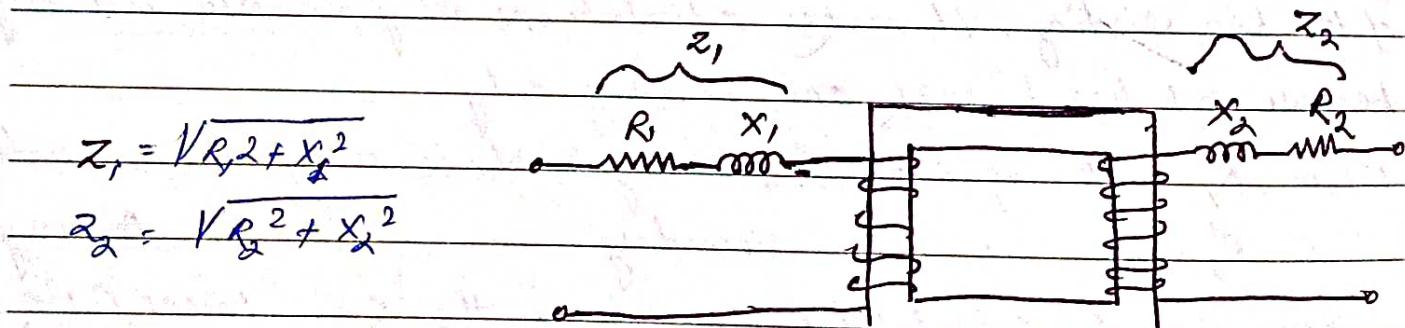


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- \* A resistance of  $R_1$  in Primary is equivalent to  $K^2 R_2$  in Secondary
- \* "  $R_2$  in Secondary "  $R_2/K^2$  in Primary
- \* But in practical ~~is~~ all the flux linked with primary
- \* does not link the secondary, it is known as primary Leakage flux
- \* This leakage flux depends on primary and secondary current. Leakage flux linking with each winding produces a self induced emf in that winding. Hence its effect is equivalent to a small inductive coil in series with each winding.

→ The primary voltage  $V_1$  will have to supply reactive drops  $jX_1$  in addition to  $jR_1$  and  $I_2$  have to supply  $jR_2$  and  $jX_2$



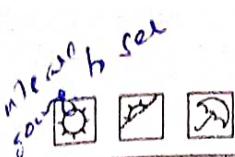
$$V_1 = E_1 + j(R_1 + jX_1) \quad \& \quad E_2 = V_2 + j(R_2 + jX_2)$$

$$\Rightarrow V_1 = E_1 + j_1 Z_1 \quad \Rightarrow E_2 = V_2 + j_2 Z_2$$

Leakage reactances can also be transferred from one winding to the other in the same way as resistance

$$X'_2 = X_2/K^2 \quad \& \quad X'_1 = K^2 X_1$$

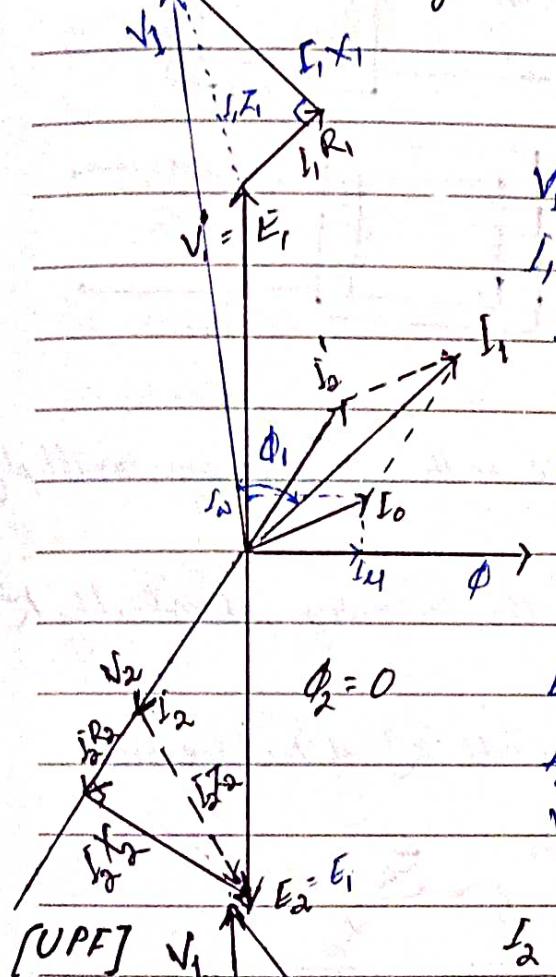
$$\text{And } X_{01} = X_1 + X'_2 = X_1 + X_2/K^2 \quad \text{and} \quad X_{02} = X_2 + X'_1 = X_2 + K^2 X_1$$



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## \* Phasor Diagram of T/F (with winding resistances, leakage reactance) And with UPF, Lagging P.F & Leading P.F



$V_1$  = Voltage drop in the primary winding

$I_1$  = Primary current

$I_0$  = Primary No load current

$I_1'$  = Load component of Primary current

$E_1$  = Induced emf (Primary)

$I_{0M}$  = Magnetising component of No load current

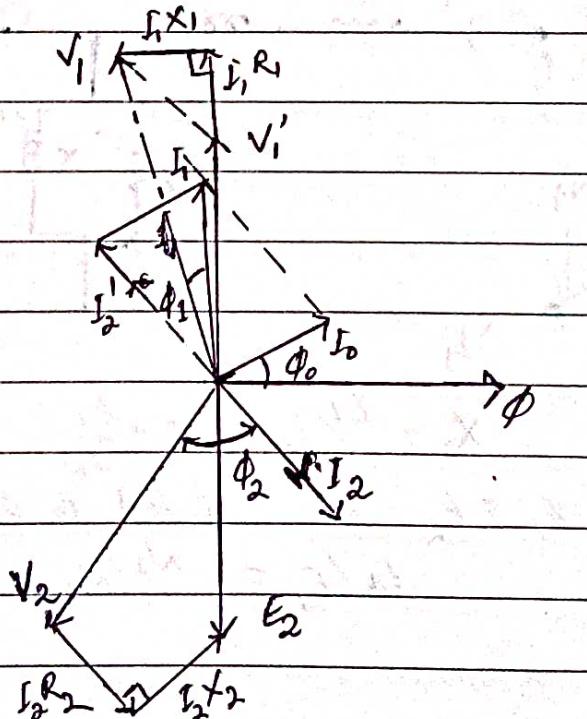
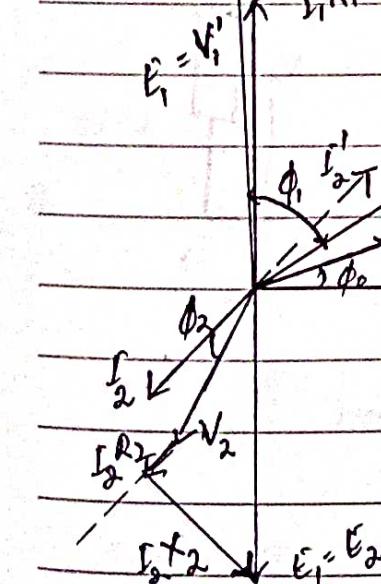
$I_{1A}$  = Active or working " "

$\phi_1 = 0$   $E_2 = E_1$   $I_2 = I_2'$   $\phi_2 = 0$

$I_2$  = Secondary current

$V_2$  = Secondary terminal voltage.

[UPF]



[Lagging P.F]

[Leading P.F]

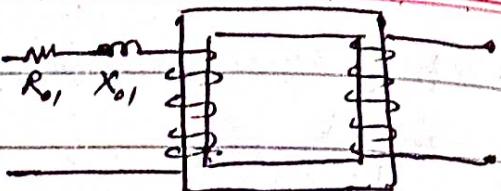


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The total impedance of transformer



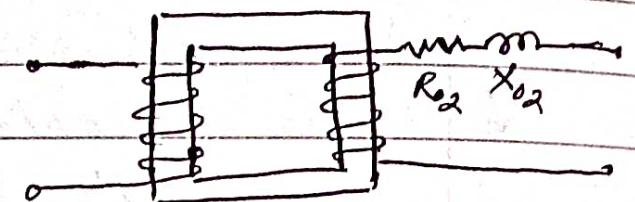
referred to Primary & Secondary is ~

$$Z_{eq} = \sqrt{R_p^2 + X_p^2}$$

$$\therefore Z_{eq} = \sqrt{R_p^2 + X_p^2}$$

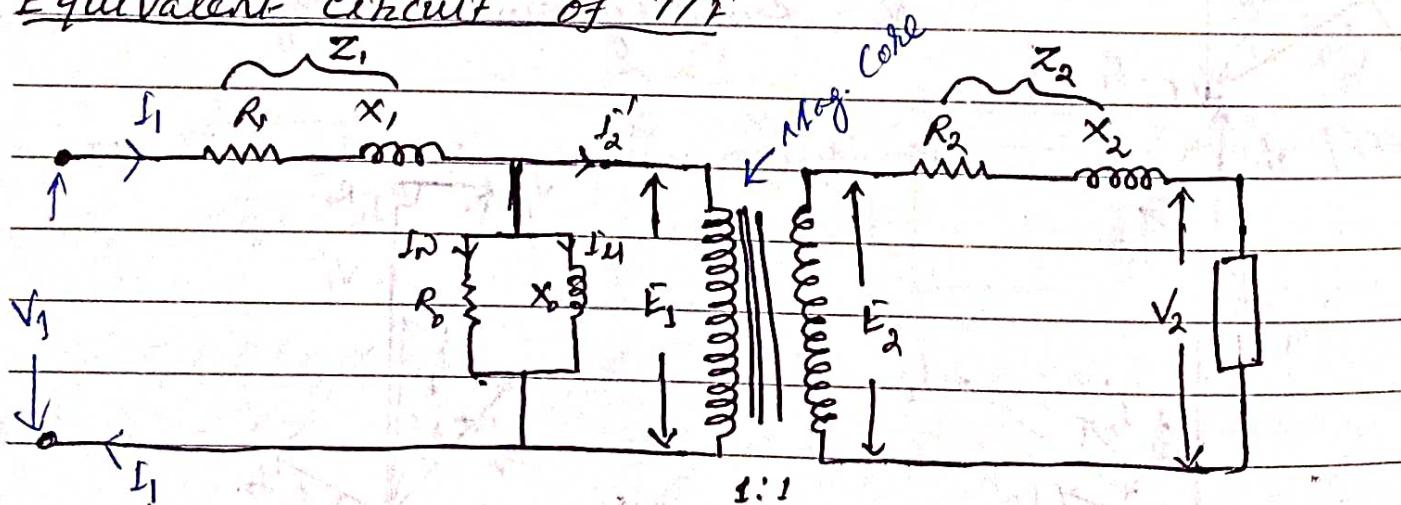
$$R_{eq} = R_p + R'_p \quad R_{eq} = R_s + R'_s$$

$$X_{eq} = X_p + X'_p \quad X_{eq} = X_s + X'_s$$



- When shifting any Resistance & Reactance to the Secondary, multiply it by  $k^2$  i.e (Transformation Ratio) $^2$
- When shifting any resistance or reactance to the Primary, divide by  $k^2$
- When shifting Voltage use 'k' only
- When shifting current (I) Primary to Secondary divide 'k' & vice versa

Equivalent circuit of T/F



$$X_o = \frac{E_1}{I_0}, \quad R_o = \frac{E_1}{I_{0D}} \quad \text{against} \quad X_o = \frac{V_1}{I_{0D}}$$

$E_1$  &  $E_2$  are related to each other by expression

$$\frac{E_2}{E_1} = \frac{N_2}{N_1} = k$$



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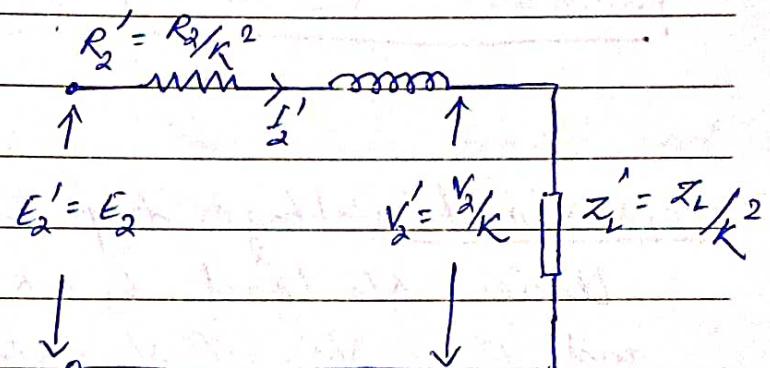
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To make transformer calculations simpler, it is preferable to transfer voltage, current & impedance.

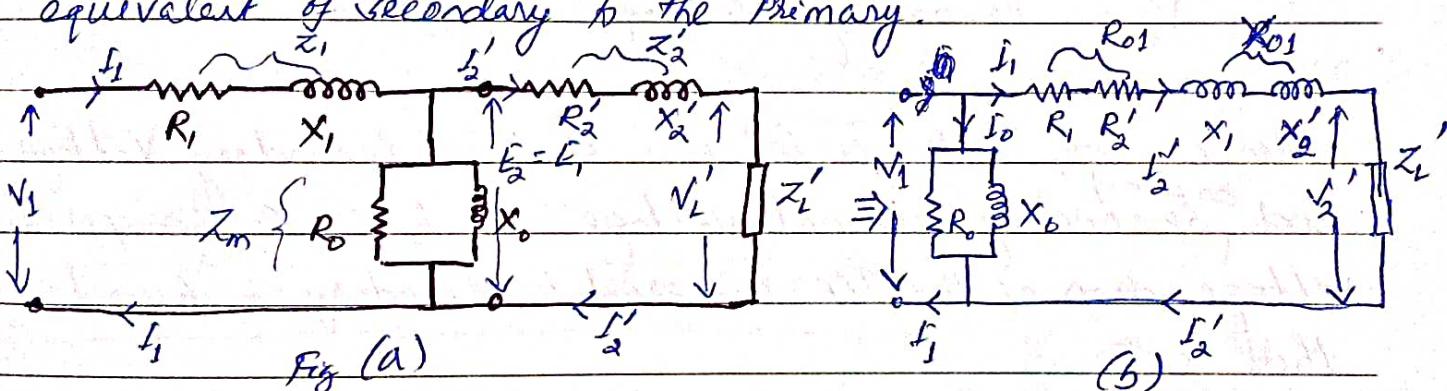
$$\text{ie } E_2' = E_2/k = E_1 \quad (\text{as } k=1)$$

$$V_2' = V_2/k, \quad I_2' = kI_2, \quad R_2' = R_2/k^2, \quad X_2' = X_2/k^2 \quad \text{so } Z_2' = R_2'/k^2$$

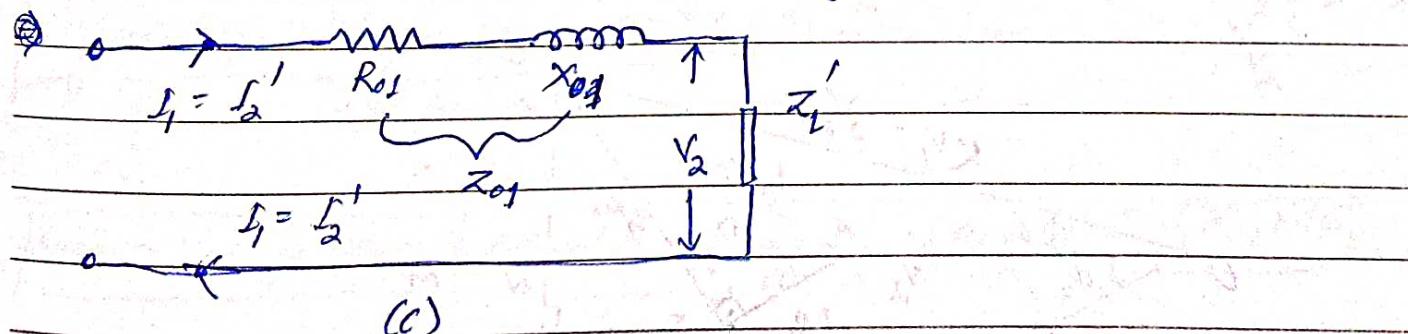
So primary equivalent of  
secondary circuit can  
be drawn as →



Total Equivalent circuit of the T/F is obtained by Primary equivalent of secondary to the Primary.



At Last, the circuit is simplified by omitting  $I_0$



From Fig (c) the equivalent impedance between input is

~~$$Z = Z_1 + Z_m || (Z_2' + Z_L')$$~~

From

$$\text{i.e } Z = Z_1 + \frac{Z_m (Z_2' + Z_L')}{Z_m + Z_2' + Z_L'}$$

$Z_m$  = Impedance of Exciting circuit

$$\therefore V_1 = I_1 \left[ Z_1 + \frac{Z_m (Z_2' + Z_L')}{Z_m + Z_2' + Z_L'} \right]$$

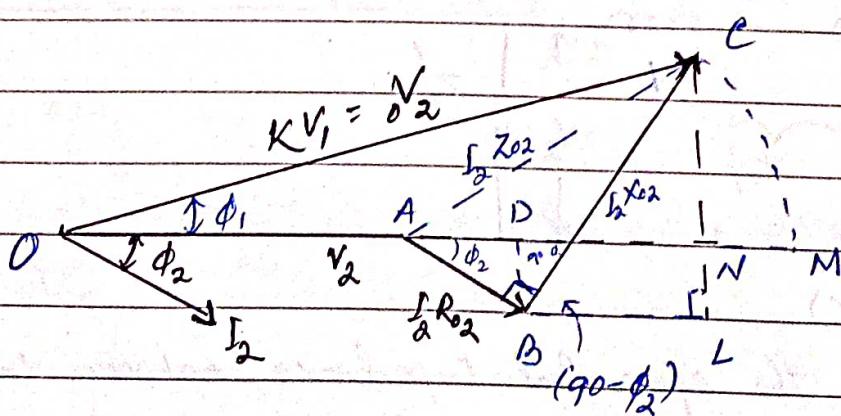
### Approximate voltage drop in a T/F

Under No Load  $E_2 = KE_1 = KV_1$  ( $V_1$  is approximately equal to  $E_1$ ) and  $E_2 = oV_2$  ( $oV_2$  = No Load Secondary terminal voltage)

$$\therefore oV_2 = KV_1$$

$V_2$  = secondary voltage on load.

Hence  $\Delta V_2$ , difference between no load secondary voltage and  $\text{on load}$  secondary terminal voltage is  $I_2 Z_{o2}$ . The approximate voltage drop of the T/F referred to secondary is found thus:



(Fig A)



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From Fig A' total voltage drop  $\int_2 Z_{02} = AC = AN$

which is approximately equal to  $AN$

To calculate approximate voltage drop  $AN'$ ,

i) Perpendicular line  $f$  on  $AN$  is drawn from point 'B'

ii) " " " on  $AN$  from C and is extended downwards.

iii) Parallel line wrt  $AN$  from point B is drawn and it intersect extended  $CN$  to point 'L'

Now,

Approximate voltage drop ( $CAN$ ) =  $AD + DN$

$$= \int_2 R_{02} \cos \phi + \int_2 X_{02} \sin \phi$$

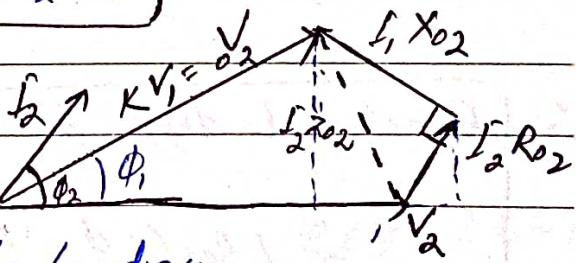
Again where  $\phi = \phi_2 = \phi$  (approx)

So  $\boxed{AN = \int_2 R_{02} \cos \phi + \int_2 X_{02} \sin \phi}$  for lagging P.F

Similarly for Leading P.F

Approximate voltage drop =  $\int_2 R_{02} \cos \phi - \int_2 X_{02} \sin \phi$

In general app. Vol. drop =  $\int_2 R_{02} \cos \phi \pm \int_2 X_{02} \sin \phi$   
+ve for lagging & -ve for leading



Again Voltage drop referred to primary =  $\int_1 R_0 \cos \phi \pm \int_1 X_{01} \sin \phi$



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## Exact Voltage Drop

From fig A' exact voltage drop can be calculated by adding NM to AN.

Considering right angle at OCN

$$NC^2 = OC^2 - ON^2 = (OC+ON)(OC-ON) = (OC+ON)(OM-ON)$$

$$\Rightarrow NC^2 = 2OC \times NM$$

{as OC = OM (radius)}

and OC ≈ OM so OC+ON = 2OC

$$\Rightarrow NM = NC^2 / 2OC \quad \text{Now } NC = LC - LN = LC - BD$$

$$\therefore NC = I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi$$

$$\text{And } NM = \frac{(I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi)^2}{2 V_2}$$

∴ Exact voltage drop for lagging p.f. is

$$= (I_2 R_{02} \cos \phi + I_2 X_{02} \sin \phi) + \frac{(I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi)^2}{2 V_2^2}$$

∴ Exact Vol. drop for leading p.f. is

$$= I_2 R_{02} \cos \phi - I_2 X_{02} \sin \phi + \frac{(I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi)^2}{2 V_2^2}$$

In general, voltage drop is  $= (I_2 R_{02} \cos \phi + I_2 X_{02} \sin \phi) + \frac{(I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi)^2}{2 V_2^2}$

$$\% \text{ Drop}^2 = \frac{I_2 R_{02} \cos \phi \pm I_2 X_{02} \sin \phi}{V_2} \times 100 + \frac{(I_2 X_{02} \cos \phi - I_2 R_{02} \sin \phi)^2}{2 V_2^2}$$



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## Regulation of T/F

When a T/F is loaded with a constant primary voltage, the secondary voltage decreases for lagging PF and increases for leading PF because of its internal resistance and leakage reactance.

Let  $\emptyset V_2$  = No Load Secondary Terminal Voltage  
=  $E_2 = E_k = KV_1$  because impedance drop is negligible at no-load.

$V_2$  = Secondary terminal voltage on full load.

Regulation of T/F is the change in secondary terminal voltage from no-load to full load i.e.  $(\emptyset V_2 - V_2)$

$$\checkmark \% \text{ Voltage Regulation (down)} = \frac{\emptyset V_2 - V_2}{\emptyset V_2} \times 100$$

$$\checkmark \% \text{ (UP)} = \frac{\emptyset V_2 - V_2}{V_2} \times 100$$

$$\text{or Approximate \% V.R.} = \frac{I_2 R_2 \cos \phi + I_2 X_2 \sin \phi}{V_1} \times 100$$

$$= \frac{I_2 R_2 \cos \phi + I_2 X_2 \sin \phi}{V_1} \times 100 + \frac{1}{2V_1} (I_2 R_2 \cos \phi - I_2 X_2 \sin \phi)^2 \times 100$$

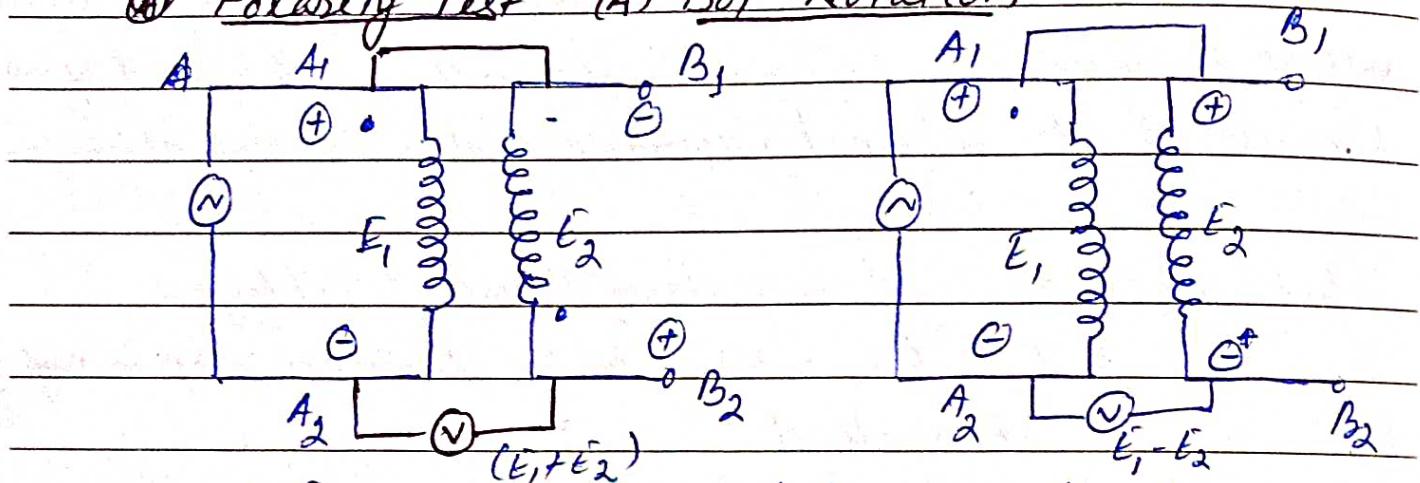


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## Testing of Transformer +

### (A) Polarity Test (A) Dot Notation



Terminal  $A'$  and  $B_1$  are shorted and Voltmeter is connected across terminal  $A_2$  and  $B_2$  and checks the voltmeter reading.

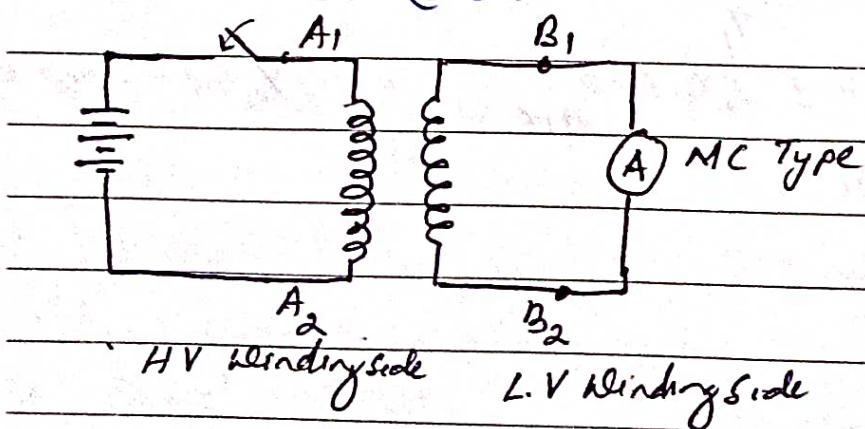
If Voltmeter Reading is

$$(i) V = E_1 + E_2 \rightarrow \text{Then series Additive polarity}$$

(ii) if  $V = E_1 - E_2$  (i.e less than  $E_1$  or  $E_2$ ) then terminal are series Subtractive.

\* Suitable for low voltage range only. (dot notation)

### (B) DC Kick Test





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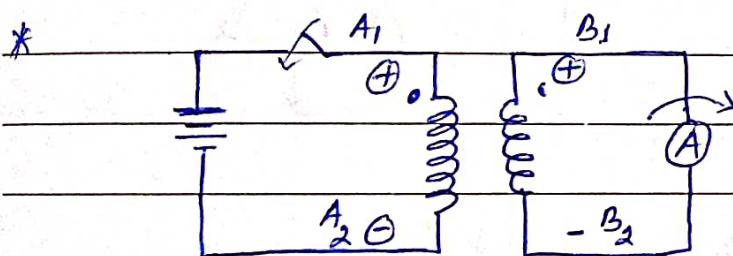
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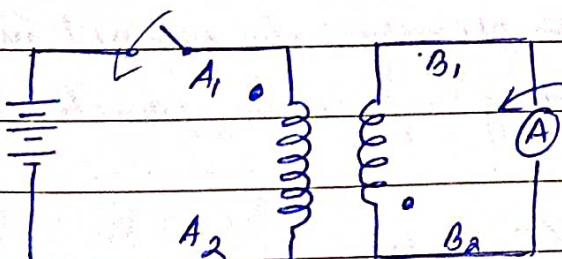
The High Voltage side winding is momentarily excited with dc supply or dc source. HV side is connected to dc source because there is a large value of resistance. And ammeter is connected at low resistance scale.

As switch is closed or during excitation a small amount of current flows in the secondary. Observe the ammeter deflection.

If Forward kick then secondary terminals are indicated with same polarities.



\* If Backward kick then secondary terminals are indicated with opposite polarities.





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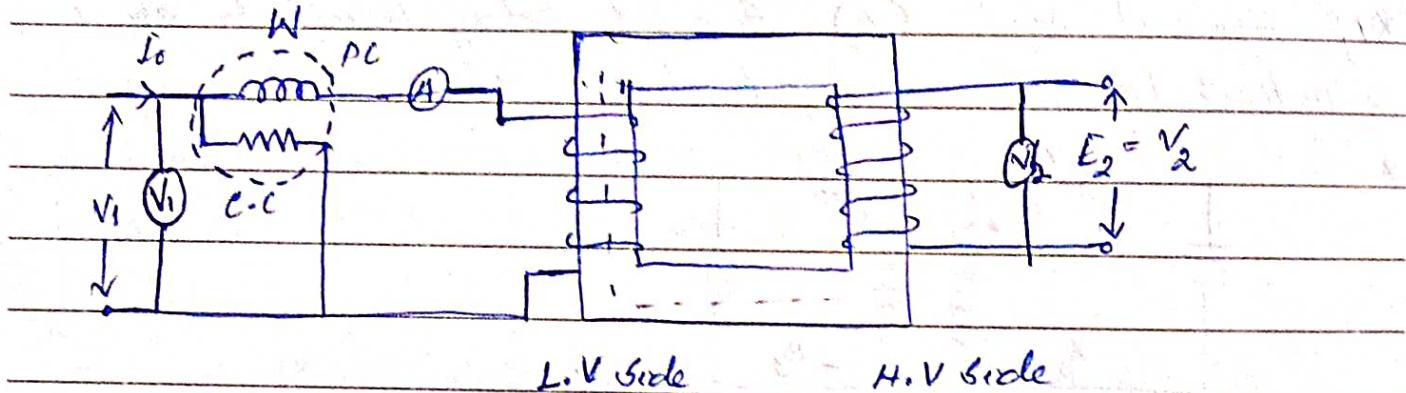
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## Open circuit test ~

The purpose of open ckt. test is to determine the no-load current ( $I_0$ ) and losses of transformer and because of which their no-load parameters  $R_0$  and  $X_0$  are determined.

→ This test is performed on Low Voltage side of Transformer.

\* and high voltage winding is left open and other side is connected to nominal supply. (measured if current is very high, which cannot be measured)



Wattmeter, ammeter and Voltmeter are connected to their L.V. side as shown in figure above.

The value of  $I_0$  is very small (ie 2-to 10% of rated Load current) so core loss negligibly small in primary side and nil in secondary side. Hence the Wattmeter reading represents practically the core loss under no-load.

$$(W) \xrightarrow{\text{Read}} \text{Core loss of T.F} = W.$$

→ As  $I_0$  is very small, the pressure coil of wattmeter and voltmeter are connected such that the current doesn't pass through the current coil of wattmeter.

→ High resistance voltmeter is connected across the secondary

$$(V) \rightarrow \text{React Induced emf in secondary side} = E_2$$



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$\textcircled{A} \rightarrow$  Reads No. load current  $I_o$

$$\text{As. } W_o = V_i I_o \cos \phi \Rightarrow \cos \phi = \frac{W_o}{V_i I_o}$$

$$\therefore I_u = I_o \sin \phi, I_w = I_o \cos \phi$$

$$\boxed{X_o = \frac{V_i}{I_M}} \quad \text{and} \quad \boxed{R_o = \frac{V_i}{I_w}}$$

As  $I_o \approx I_u$  (As current is all exciting current at no-load) and voltage drop is very small.

$$\text{Exciting admittance } (Y_o) = \frac{I_o}{V_i}$$

$$\text{conductance } (G_o) \text{ is given by } W_o = V_i^2 G_o \Rightarrow \boxed{G_o = \frac{W_o}{V_i^2}}$$

$$\text{Exciting susceptance } (B_o = \sqrt{(Y_o)^2 - (G_o)^2})$$

Core loss depends on the frequency and maximum flux density when volume and the thickness of core laminations are given.  $f$  const.

Core loss = Hysteresis Loss ( $W_h$ ) + Eddy current Loss ( $W_e$ )

$$\Rightarrow W_h = n B_{max}^2 f v + K B_{max}^2 f^2 t^2$$

$$\Rightarrow W_e = A f + B f^2$$

$$\Rightarrow \boxed{\frac{W_e}{f} = A + B f}$$



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## Short-Circuit or Impedance Test.

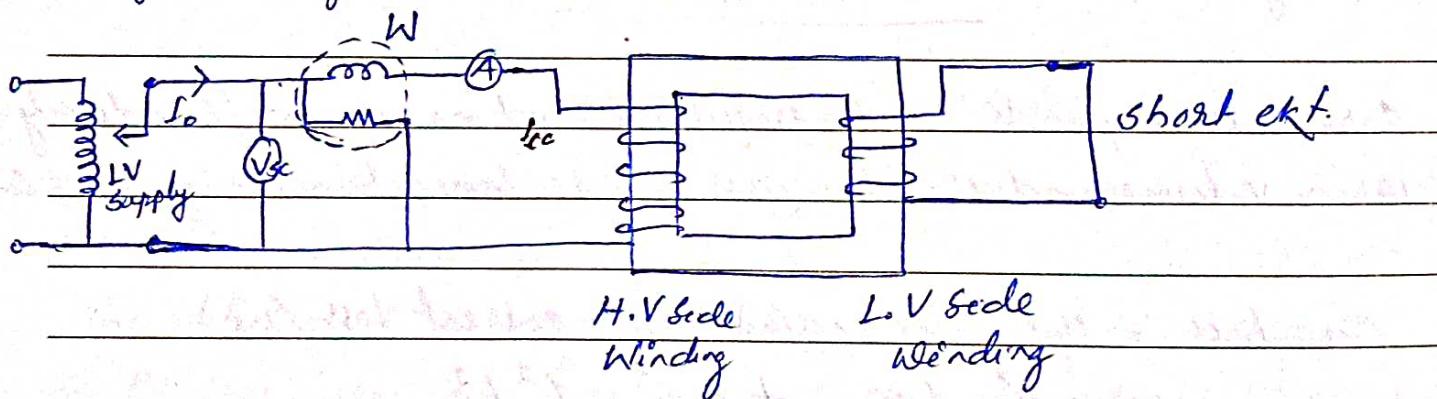
The economical method for determining

i) Equivalent impedance ( $Z_1$  or  $Z_2$ ) Leakage reactance ( $X_1$  or  $X_2$ ) and total resistance ( $R_1$  or  $R_2$ )

ii) To find variable loss i.e. core loss at full load and Efficiency

→ This test is performed on the secondary winding side i.e. High Voltage side.

Wattmeter, Voltmeter and ammeter are connected to the high voltage side as shown below.



A low voltage (usually 5 to 10% of normal primary voltage) at correct frequency is applied and is continuously increased till full load current are flows in both primary & secondary side indicated by respective Ammeter.

As applied voltage is small percentage of normal voltage, the mutual flux  $\phi$  is also small and hence core loss is very small. The result that the wattmeter



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reading represent the full load cu. loss or  $I^2R$  loss of whole T/F. Equivalent circuit under short ckt. can be drawn as fig.c

$W_{sc} = \text{Wattmeter Reading or } I^2R \text{ loss}$        $\uparrow I_1 = I_2' \quad R_{o1} \quad X_{o1} \uparrow$

$V_{sc} = \text{Voltmeter Reading or Voltage } V_1$        $V_2' = 0$

required to circulate rated load

current

$I_{sc}$  = Ammeter Reading or Rated full load current

$$\text{As } W_{sc} = I_{sc}^2 R_{o1} \Rightarrow \left| \frac{R_{o2}}{R_{o1}} = \frac{W_{sc}}{I_{sc}^2} \right| = \frac{\text{Wattmeter Reading}}{(\text{Ammeter Reading})^2}$$

$$V_{sc} = I_{sc} \times Z_{o1} \Rightarrow Z_{o2} = \frac{V_{sc}}{I_{sc}}$$

then

$$X_{o2} = \sqrt{(Z_{o2}^2 - R_{o2}^2)} \quad R_{o1} = \left( \frac{N_1}{N_2} \right)^2 R_{o2}$$

Why Transformer Rating in KVA?

\* As cu loss of a T/F depends on current and iron loss on voltage i.e. total loss depends on Volt-ampere (VA) and not on phase angle between voltage and current. That is why rating of T/F is in KVA and not in kWatt.

\*



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## Efficiency of Transformer

The efficiency of a T/F at a particular load and power factor is defined as the output divided by the input. The two being measured in the same units (either watt or kilowatt).

$$\text{Efficiency}(\%) = \frac{\text{output (kW)}}{\text{input (kW)}} = \frac{\text{output}}{\text{output + Losses}}$$

Though T/F being efficient piece of equipment, has very small loss, hence it is impractical to try to measure efficiency by measuring input & output.

\* A better method is to determine the losses & then to calculate efficiency.

$$\eta = \frac{\text{output}}{\text{output + Losses}} = \frac{\text{output}}{\text{output + Cu. Loss + Iron Loss}}$$

$$= \frac{\text{Input - Losses}}{\text{Input}} = 1 - \frac{\text{Losses}}{\text{Input}}$$

\* Cu. Loss can be calculated from short circuit test

\* Iron loss or core loss from open circuit test.

$$\eta_{FE} = \frac{V_1 I_1 \cos \phi_1 - I^2 R_{01} - N_i}{V_1 I_1 \cos \phi_1} \quad \text{Referred to Primary}$$

$$\eta_{PT} = \frac{V_2 I_2 \cos \phi_2}{V_2 I_2 \cos \phi_2 + P_1 \text{ cu-losses + Iron Loss}} \times 100$$
$$\quad \quad \quad (I_2^2 R_{02}) (N_i)$$



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$$\eta_{x \text{ of FL}} = \frac{x \times \text{Full Load KVA} \times \text{P.f.}}{(x \times \text{full load KVA} \times \text{P.f.}) + x^2(I_2^2 R_{02}) + W_i} \times 100$$

at given P.F.

where  $x = \text{Ratio of actual to full load KVA}$

Condition for Max. Efficiency is

$$\eta = 1 - \frac{\text{Losses}}{\text{Input}} = 1 - \frac{I_1 R_{01}}{V_1 \cos \phi_1} - \frac{W_i}{V_1 I_1 \cos \phi_1}$$

for  $\eta$  to be maximum  $\frac{d\eta}{dI_1} = 0$  —

$$\therefore \frac{d\eta}{dI_1} = 0 - \frac{R_{01}}{V_1 \cos \phi_1} - \frac{W_i}{V_1 I_1^2 \cos \phi_1} = 0$$

$$\Rightarrow \frac{R_{01}}{V_1 \cos \phi_1} = \frac{W_i}{V_1 I_1^2 \cos \phi_1} \Rightarrow W_i = I_1^2 R_{01} \text{ or } I_2^2 R_{02}$$

: [cu. loss = iron loss]

To get maximum efficiency cu. loss of a T/F must be equal to core loss (iron loss) or constant loss.

All day efficiency?

All day efficiency means the power consumed by the transformer throughout the day.

$$\text{All day Efficiency (A/day)} = \frac{\text{Output in kWh}}{\text{Output + losses}} \quad (\text{for 24 hours})$$

All day efficiency can track the performance of distribution T/F



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### → Load corresponding to Max Efficiency

Let  $\chi'$  be the ratio of actual to full load at which efficiency is maximum.

We know the condition for max efficiency is  $Cu\text{.loss} = \text{Iron loss}$   
Hence.

Cu. loss at  $\chi$  of F.L = Iron Loss

$$\Rightarrow \chi^2 \frac{I^2}{2} R_{02} = Ni$$

$$\Rightarrow \chi = \sqrt{\frac{Ni}{\frac{I^2}{2} R_{02}}} \quad i.e. \quad \boxed{\chi = \sqrt{\frac{\text{Iron Loss}}{\text{FL Cu Loss}}}}$$

$$\boxed{\text{Load KVA for } \eta_{\max} = \text{FL KVA} \times \sqrt{\frac{\text{Iron Loss}}{\text{FL Cu Loss}}}}$$

### Variation of Efficiency with Power Factor

$$\eta = 1 - \frac{\text{Loss}}{\text{Input}} = 1 - \frac{\text{Losses}}{\frac{V_2 I_2 \cos \phi_2}{2} + \text{Losses}}$$

Let  $\text{Losses}/\frac{V_2 I_2}{2} = y$  then

$$\eta = 1 - \frac{y}{\cos \phi_2 + \frac{y}{\frac{V_2 I_2}{2}}} = 1 - \frac{x}{\cos \phi_2 + x}$$

$$\Rightarrow \boxed{\eta = 1 - \frac{x/\cos \phi_2}{1 + x/\cos \phi_2}}$$



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## Parallel operation of Transformer

\* If two or more transformer are connected to a same supply on the primary side and to a same load on the secondary side, then it is called as parallel operation of transformer.

### Necessity of Parallel operation is (x)eed)

- 1) Increased Load. ≈ Necessary to supply the increased load when load exceeds the capacity of existing Transformer.
- 2) Non Availability of Large T/F ≈ When a large transformer is not available to meet the total requirement of load, then two or more small T/F are connected in parallel.
- 3) Increased reliability ≈ ~~if~~ Any Bridge fault T/F can can be taken for maintenance uninterrupted to serve the load.
- 4.) Transportation is easier for small T/F ≈ If installation site is located far away, then transportation is easier for and economical for smaller units.

### conditions for parallel operation of T/F ≈

- 1) The line voltage ratio of two T/F must be equal.
- 2) The per unit impedance of each T/F must should be equal and they should have same ratio of equivalent leakage reactance to the resistance. ( $X/R$ )
- 3) The T/F should have same secondary winding polarity.
- 4) " " " " phase sequence.
- 5) The " " the zero relative phase displacement between secondary line voltage ( $3\phi$  T/F)



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## Auto - Transformer

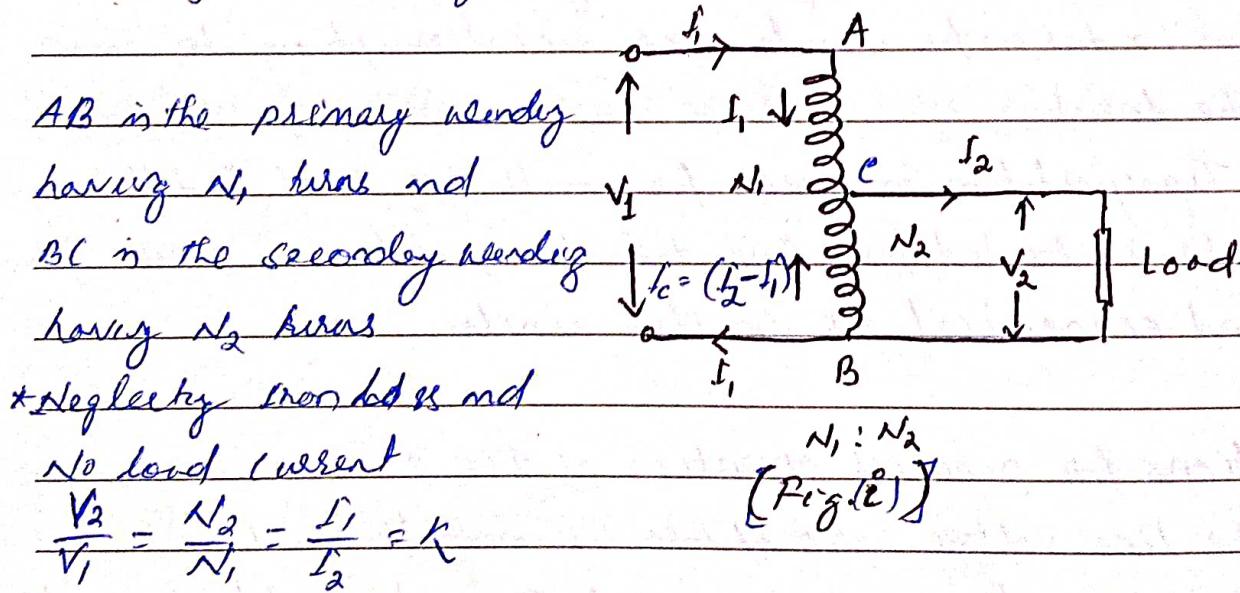
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A transformer in which a part of the winding is common to both the primary and secondary circuit is called an Auto-transformer.

→ In Auto-transformer the primary and secondary winding are not electrically isolated from each other as in case of two-winding transformer.

Principle of operation is

The principle of operation is same as that of common transformer. It works on the principle of Faraday's Law of Electromagnetic Induction.



As current in primary and secondary are practically in phase opposition and  $I_2$  is greater than  $I_1$ . (Lenz's Law)

So Common Current ( $I_c$ ) =  $I_2 - I_1$



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\* Saving of Cu. (Comparing Auto-transformer with two winding Transformer)

In case of stepdown T/F as shown in Fig-(a),  $k < 1$

Volume and hence weight of Cu. is proportional to the length and area of the cross-section of conductor

Length of conductor  $\propto$  Number of turn ( $N$ )

Cross section  $\propto$  current ( $I$ )

$\therefore$  Weight  $\propto$  Product of current and number of turns (i.e.  $NI$ )

Referring figure-(2)

Total weight of Cu. in auto T/F  $\propto$  weight of Cu. in section AC  
 $(W_a)$

+ weight of Cu. in section BC

$$\Rightarrow W_a \propto (N_1 - N_2) I_1 + N_2 (I_2 - I_1) \quad \text{eq(1)}$$

Total weight of Cu. in two winding T/F ( $W_b$ )  $\propto$  weight of Cu. in Primary side  
+ weight of Cu. in Secondary side

$$\Rightarrow W_b = N_1 I_1 + N_2 I_2 \quad \text{eq(2)}$$

from eq(1) and eq(2)

$$\therefore \text{Wt. of Cu. in auto-T/F} = (N_1 - N_2) I_1 + N_2 (I_2 - I_1)$$

$$\text{Wt. of Cu. in ordinary T/F} \quad N_1 I_1 + N_2 I_2$$

$$= \frac{N_1 I_1 + N_2 I_2 - N_2 I_1 - N_2 I_1}{N_1 I_1 + N_2 I_2} = 1 - \frac{2 N_2 I_1}{N_1 I_1 + N_2 I_2}$$

$$= 1 - \frac{2 N_2 I_1 \times \frac{1}{N_1 I_1}}{(N_1 I_1 + N_2 I_2) \frac{1}{N_1 I_1}} = 1 - \frac{2 \frac{N_2}{N_1}}{1 + \frac{N_2}{N_1} \times \frac{I_2}{I_1}}$$

$$= 1 - k \quad \left[ \text{as } \frac{N_2}{N_1} = k \text{ and } \frac{I_2}{I_1} = k \right]$$



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c.e  
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$$\left[ \frac{W_a}{W} = 1 - k \right]$$

or  $\frac{\text{wt. of Cu. in Auto T/F}}{\text{wt. of Cu. in ordinary T/F}} = 1 - k$

$\therefore \text{Weight of Cu. in Auto T/F} = (1 - k) \text{ Weight of Cu. in Ordinary Transformer}$

$$\text{Saving} = W_0 - W_a = W_0 - (1 - k) W_0$$

$\Rightarrow \text{Saving} = k W_0$

- \* Saving of cu. will increases as 'k' approaches unity (1)
- $\rightarrow$  The power transformed inductively =  $(1 - k)$  Input
- $\rightarrow$  Rest Power =  $k \times$  Input is conducted directly from the source to load i.e it is transferred conductively to the load.

### Uses of Auto-transformer:

- 1) It is used as a saturator to give 50 to 60% of full voltage to starter of a squirrel cage induction motor during starting.
- 2) It is used to give a small boost to a distribution cable to correct the voltage drop.
- 3) It is used as a Voltage Regulator.
- 4) Used as interconnecting T/F in 132 KV/330 KV power system ratios are
- 5) Used in control equipment for 1φ and 3φ electrical locomotives.



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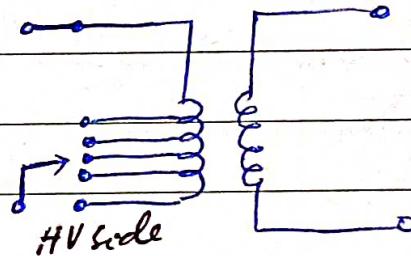
## Tap changer with Transformer.

The change of voltage is affected by changing the number of the transformer provided with taps. For efficiently close control of voltage, tap are usually provided on the high voltage winding of the T/F.

### Type of tap-changing T/F

- Off-Load tap changing T/F
- On-Load tap changing T/F
- OFF-Load tap changing T/F

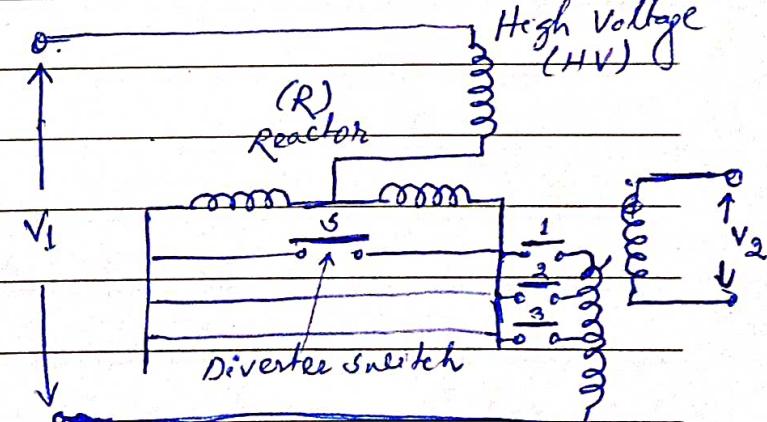
In this method, the transformer is disconnected from main supply when the tap setting is to be changed. The tap setting is done manually.



[OFF-Load tap changing T/F]

### b) On-Load tap changing T/F

on load tap changing is used to supply uninterrupted power supply. While tapping, two essential conditions are to be filled



(i) The Load circuit should not

be broken to avoid arcing and prevent the damage of contacts.

(ii) No part of the winding should be short-circuited while adjusting the tap.



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The transformer is operated with switch-1 and 3' closed. To change to tap 2, switch 3' is opened and 2 is closed. Switch 1 is then opened, and 3' is closed to complete the tap change. It is to be noted that the diverter switch operates on load and no current flows through selector switch 3' during tap changing. During the tap change only half of the reactance which limits the current is connected in the circuit.

# THREE PHASE TRANSFORMER



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Even though most of the equipments are connected by the single phase transformer, but these are not preferred for large power distribution in aspect of economy. The 3 $\phi$ -PTO power ~~are~~ is <sup>used in</sup> almost all field of electrical power system such as power generation, transmission and distribution. Therefore 3 $\phi$  T/F is used to step-up or step-down the voltage in 3 $\phi$  system.

- 3 $\phi$  T/F is smaller and lighter in construction for the same power handling capacity and better operating characteristics.
- In order to reduce the power loss to the distribution end, the power is transmitted at higher voltage (132kV or 400kV) with 3 $\phi$  T/F and again to step down at transmission and distribution end to level of 6600V, 400V, 230V, etc.
- Due to single 3 $\phi$  T/F amount of iron in the core and insulation materials are saved.

Construction :-

A 3 $\phi$  T/F can be constructed by using common magnetic core for both primary & secondary winding. This can be of a bank of three core type T/F (3-core type single phase T/F are combined). Similarly a bank of 3 $\phi$  shell type T/F (3 shell type single phase are combined).



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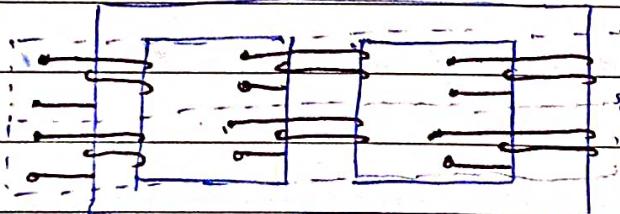
### a) Core type

Core of  $3\phi$  T/F is made up of three limbs or leg & two yokes. Magnetic path is formed between these yokes and limbs. On each limb both primary and secondary windings are wound concentrically.

- Circular cylindrical cores are used as the winding.
- Both winding (primary & secondary) of one phase are all wound on one leg.
- Under balanced condition, the magnetic flux in each phase of the leg adds up to zero. Therefore, under normal condition, no return path (leg) is needed.

But in unbalanced load, high circulating current flows and hence it may be best to use three single phase

T/F.      phase R      phase Y      phase B



[core type T/F]

### (B) Shell type

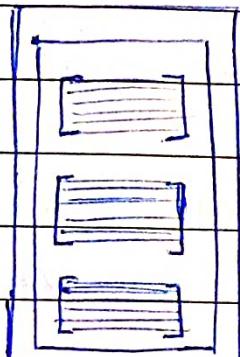
In shell type,  $3\phi$  are more independent because each phase has independent magnetic circuit compared with core type T/F. It is same as single phase shell type T/F. The magnetic circuit are in parallel. Due to this, the saturation effects in



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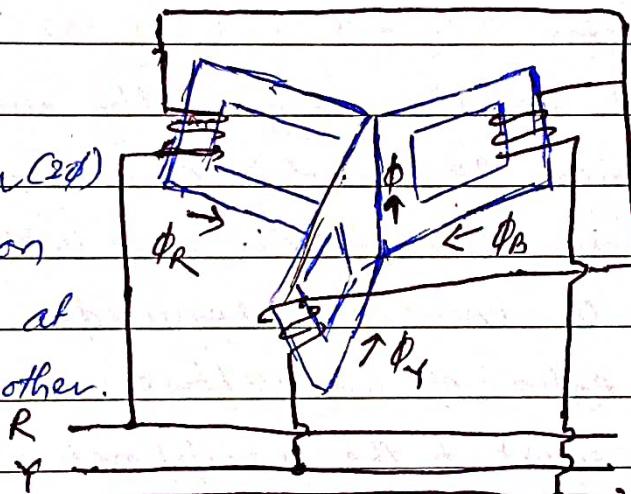
common magnetic paths are neglected. However shell type constructed T/F are rarely used in practice



Shell type

Working :-

Consider a transformer ( $2\phi$ ) is connected in star fashion. The 3 core are arranged at an angle of  $120^\circ$  to each other.



When primary is excited with  $\omega_0$  supply source, the current  $i_R, i_Y, i_B$  starts flowing individual phase winding. These produce fluxes  $\phi_R, \phi_Y$  and  $\phi_B$ . Since centre leg is common, the sum of all three core flux are carried by it. But in three phase balance system the sum of all the current is zero, so flux is also zero. Hence the centre leg doesn't carry any flux at any instant. So on removal of centre leg, it makes no difference in other condition of T/F.

So if centre leg of T/F is removed, then any two



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conductors <sup>or leg</sup> acts as return path of the flux for third leg.

### 3 $\phi$ T/F connection

On the basic winding of a three phase T/F can be connected in various configurations as

- i) Star - Star ( $Y-Y$ )
- ii) Delta - Delta ( $\Delta-\Delta$ )
- iii) Star - Delta
- iv) Delta - Star
- v) Open delta and vi) Scott connection

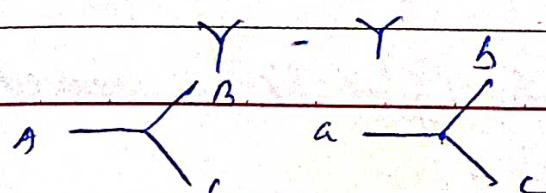
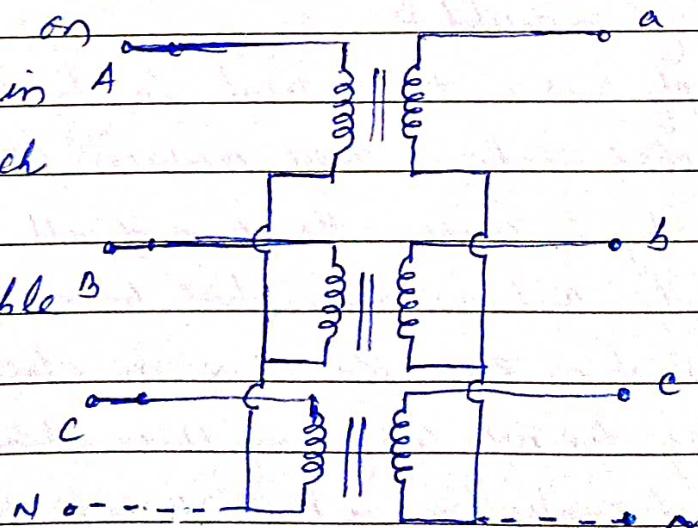
#### (2) Star - Star ( $Y-Y$ )

•  $Y-Y$  connection is generally used for small, high-voltage transformer. Because of  $Y$  connection, number of required turns/phase is reduced (as phase voltage is  $\frac{1}{\sqrt{3}}$  times of line voltage) so insulation required is also reduced.

• The ratio of line voltage on the primary and secondary side is equal to the transformation ratio of the transformer.

• Line voltage on both sides are in a phase with each other.

• This is suitable for balanced load only.





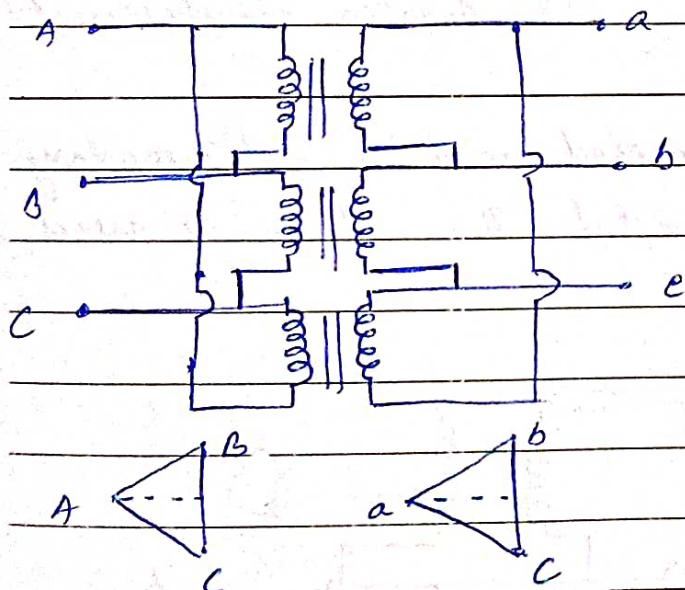
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## (ii) Delta - Delta (Δ-Δ)

- This connection is generally used for large, low-voltage transformers. Number of turns/phase is relatively greater than that for star-star connection.
- The ratio of line voltage on primary and secondary side is equal to the transformation ratio.
- This can be used even for unbalanced loading.
- This can be operated in open delta connection, if one transformer is disable, but with reduced capacity.



## (iii) Star - Delta (Y-Δ)

- The primary winding is star (Y) with ground neutral and secondary winding is delta connected.
- This is used in step-down transformer at the substation end of the transmission line.



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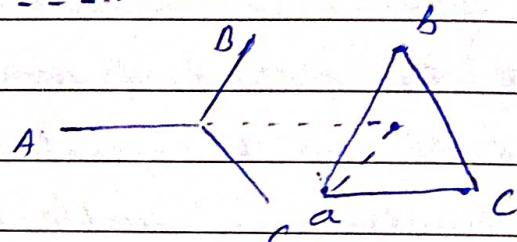
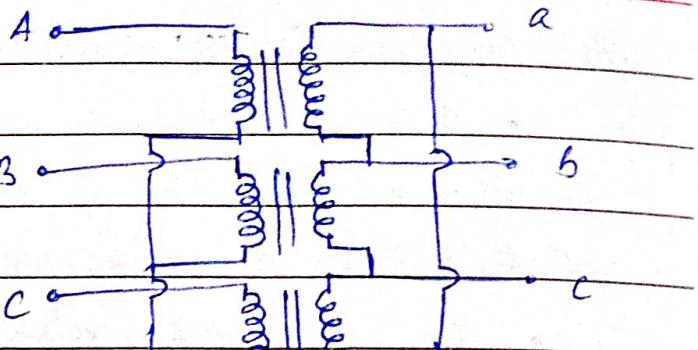
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- The ratio of Secondary to primary line voltage is  $\sqrt{3}$  times of transformation ratio.

- There is  $30^\circ$  shift between the primary and secondary line voltages.



( $30^\circ$  Angular displacement)

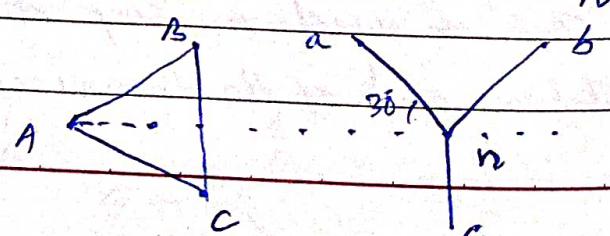
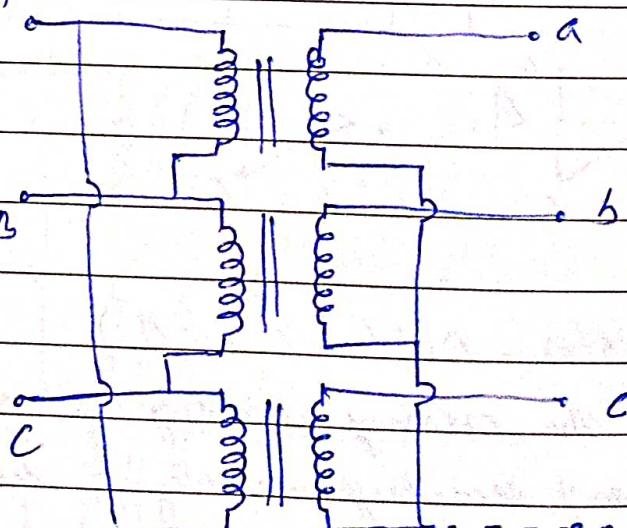
## (iv) Delta-star ( $\Delta-Y$ )

- The primary winding is connected in delta and secondary is star with neutral grounded. Thus it can be used to  $3\phi$  4-wire service.

- This type of connection is mainly used in step-up transformer at the beginning of transmission line.

- Ratio of secondary to primary line voltage is  $\sqrt{3}$  times the transformation ratio.

- There is  $30^\circ$  shift between primary & secondary line voltage.





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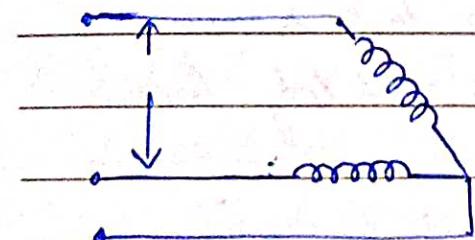
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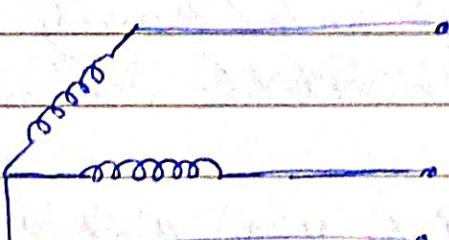
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## ④ Open Delta

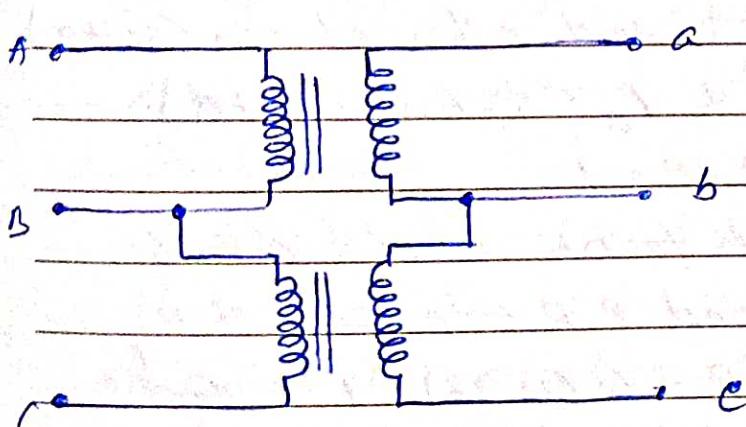
Open delta transformer uses only two windings of primary and secondary and transform a 3<sup>ph</sup> voltage. It is not used commonly. Typically used for small load where cost is important. Alternatively they could be used as an emergency measure, should one winding only of a transformer fail.



(Primary Connection)



(Secondary Connection)



Power delivered  $\approx$

in delta connection

$$I_L \text{ (Line current)} = \sqrt{3} I_{ph}$$

$$V_L = V_{ph}$$

$$\text{Power } 0/p = \sqrt{3} V_L I_L$$

$$= \sqrt{3} V_L V_{ph} I_{ph}$$

open delta

$$I_L = I_{ph}$$

$$V_L = V_{ph}$$

$$\text{Power } 0/p = \sqrt{3} V_L I_L$$

$$= \sqrt{3} V_L I_{ph}$$



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$$\begin{aligned} V-V \text{ capacity} &= \frac{\sqrt{3} V_L \text{ fph}}{3} = 0.577 \text{ or } 57.7\% \\ \Delta-\Delta \text{ capacity} &= \frac{3 V_L \text{ fph}}{3} \end{aligned}$$

\* open delta will only deliver 57.7% of the power of conventional T/F (not 66.7% as may be expected)

OR

only 86.6% of rated capacity of the two remaining transformer is available. It is the reliability factor

Power supplied by V-V Bank =

When a V-V bank of two T/F supplies a balanced 3-Ø load of power factor  $\cos(\alpha)$  then one T/F operates at a P.F of  $\cos(30-\alpha)$  and other at  $\cos(30+\alpha)$ .

$$P_1 = KVA \cos(30^\circ - \alpha) \text{ & } P_2 = KVA \cos(30 + \alpha)$$

1) \* When  $\alpha = 0^\circ$  i.e. load P.F = 1

Each T/F has power factor  $\cos 30^\circ = 0.866$

\* (ii) When  $\alpha = 30^\circ$  i.e. load P.F = 0.866

One T/F will have P.F =  $\cos(30-30) = 1$  and  
other .. .. .. =  $\cos(30+30) = 0.866$

\* (iii) When  $\alpha = 60^\circ$  i.e. load P.F =  $\cos 60^\circ = 0.5$

One T/F will have P.F =  $\cos(30-60) = 0.866$  &  
other .. .. .. =  $\cos(30+60) = 0$

✓ i.e. second T/F will not supply any load and first one will supply entire load.



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### use of open delta connection :-

- used for small 3- $\phi$  load.
- used where the current load is less and may increase in future
- There is no shifting in output, so can be used for unbalanced load and where harmonics are involved.

### Parallel operation of 3 $\phi$ Transformer :-

#### Conditions

- i) The line voltage ratio of the transformer must be same.
  - ii) The transformers should have equal per unit leakage impedance
  - iii) The ratio of equivalent leakage reactance to equivalent resistance should be same for all the transformers.
  - iv) The transformers should have the same polarity.
- Along with these four conditions which are applicable for parallel operation of 1 $\phi$  T/F, two more condition should be fulfilled for parallel operation of 3 $\phi$  T/F.

- (i) The relative phase displacement between the secondary line voltage of all transformers should be zero. This means that transformers to be connected in parallel must belongs to same group like Yy0 and Dd0.



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- vi) The phase sequence of secondary line voltages of all the transformers should be same

### Maintenance of Power Transformer

Maintenance is done at regular interval for trouble free operation and to prevent breakdown of transformer from service.

There is a check list of maintenance point being carried out as per time duration of daily, monthly, 3 month, 6 months, Yearly, once in 2 years and once in 7 to 10 years.

#### a) Daily check point

1) Check oil level of main conservator and OLTC tank. It should be as per specified level.

(B) 2) Check oil leakage from any point of T/F.

3) Check oil level in bushing, core and temperature of oil and winding whether they are normal or not

#### (B) Monthly check Point

I) Check oil level in oil cap under silica gel breather and breathing holes in silica gel



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breather and clean properly if required.

silica gel breather is blue in colour if its colour is pink then replacement or heating of silica gel is required.

2) Check oil level in Buchholz relay.

c) Check point for 3 months interval is

- 1) Check for dust deposition on bushing and tightness of oil filling plug and examine the crack in porcelain discs.
- 2) Check for cooling fan and lubricate fan bearing if required.

(d) Yearly check point is

- 1) Check the insulation of Transformer Oil.
- 2) Check mechanical factors of Buchholz relay.
- 3) Check resistive value of earth connection ( $< 1\text{ ohm}$ )



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Q1. For a 5kVA 400/200V, 50Hz, 7TF  
 open circuit test: 400V, 1A, 60Watt  $\left( R_o, X_o \text{ in } \text{Primary} \right)$   
 short circuit test: 15V, 12.5A, 50Watt  $\left( \text{Primary} \right)$   
 find (i)  $R_o$  &  $X_o$ .  $\left( \text{High Voltage} \right)$

(i) Equivalent winding resistance & leakage reactance referred to primary.

(ii) Iron & copper losses at full load

(iii) Efficiency at full load with PF at 0.8

(iv) Regulation at full load with PF at 0.8

Ans OC test-

$V_{oc} = V_i = \text{Voltmeter reading}$   $I_o = \frac{\text{Ammeter reading}}{\text{No load current}}$   $\text{Iron loss } (E_{oc}) = \text{ammeter reading}$

$V_{oc} = V_i = 400V$ ,  $I_o = 1A$ ,  $P_i = 60\text{Watt} = V_i I_o \cos \phi$

$$\therefore I_o \cos \phi \quad (\text{useful component of } I_o) = \frac{P_i}{V_{oc}} \quad \text{or } I_o \quad \text{No load current}$$

$$\Rightarrow I_o = \frac{60}{400} = 0.15 A$$

$$I_m = \sqrt{I_o^2 - I_n^2} = 0.988 A \quad (\text{Magnetizing component})$$

$$R_o = \frac{V_{oc}}{I_n} = \frac{400}{0.15} = 2666.66 \Omega$$

$$X_o = \frac{V_{oc}}{I_m} = \frac{400}{0.988} = 404.8 \Omega$$



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## short circuit (Secondary)

$$V_2 = 15 \text{ V}, I_{sc} = 12.5 \text{ A}, P_{sc} = 50 \text{ Watt}$$

$$R_{22} = \frac{P_{sc}}{I_{sc}^2} = \frac{50}{12.5^2} = 0.32 \Omega \quad \frac{I_{sc}}{K_2} = \frac{12.5}{15} = \frac{5}{6} =$$

$$Z_{22} = \frac{V_{sc}}{I_{sc}} = \frac{15}{12.5} = 1.2 \Omega$$

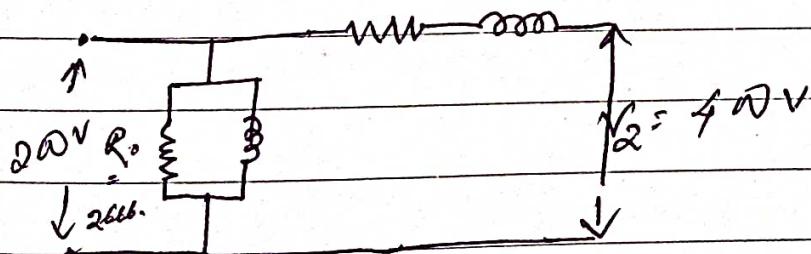
$$X_{22} = \sqrt{Z_{22}^2 - R_{22}^2} = \sqrt{1.34} = 1.156 \Omega$$

referred to Primary

$$R_{01} = \left(\frac{N_1}{N_2}\right)^2 R_{22} \text{ or } R_{01} = K^2 \times R_{22} = \frac{1}{4} \times 0.32$$

$$R_{01} = 0.8 \Omega$$

$$R_{01} = \frac{1}{4} \times 1.56 = 0.39 \Omega$$

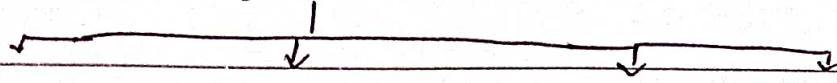




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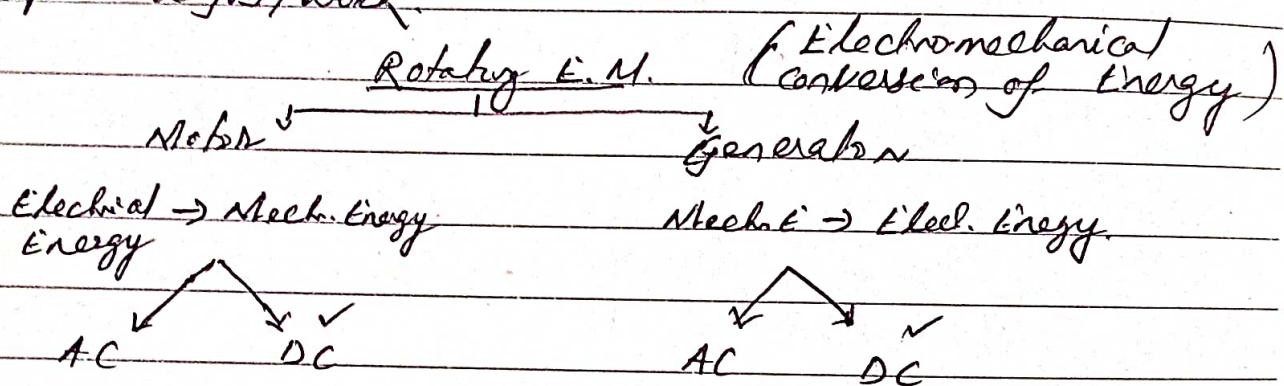
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## DC Machine



Machine      DC Machine      Advantages Application.  
Static      DC Machine      Linear operating principle  
M.F.S      Rotating (Electromechanical conversion principle)

Machine is the device which reduces our effort to accomplish a job/work.



Depending upon the type of winding the DC machine both Motor & generator are divided into

DC Machine  
Self Excited      Separately Excited DC.M  
DC M/c      Permanent Mag. DC Machine

## Application & Advantages

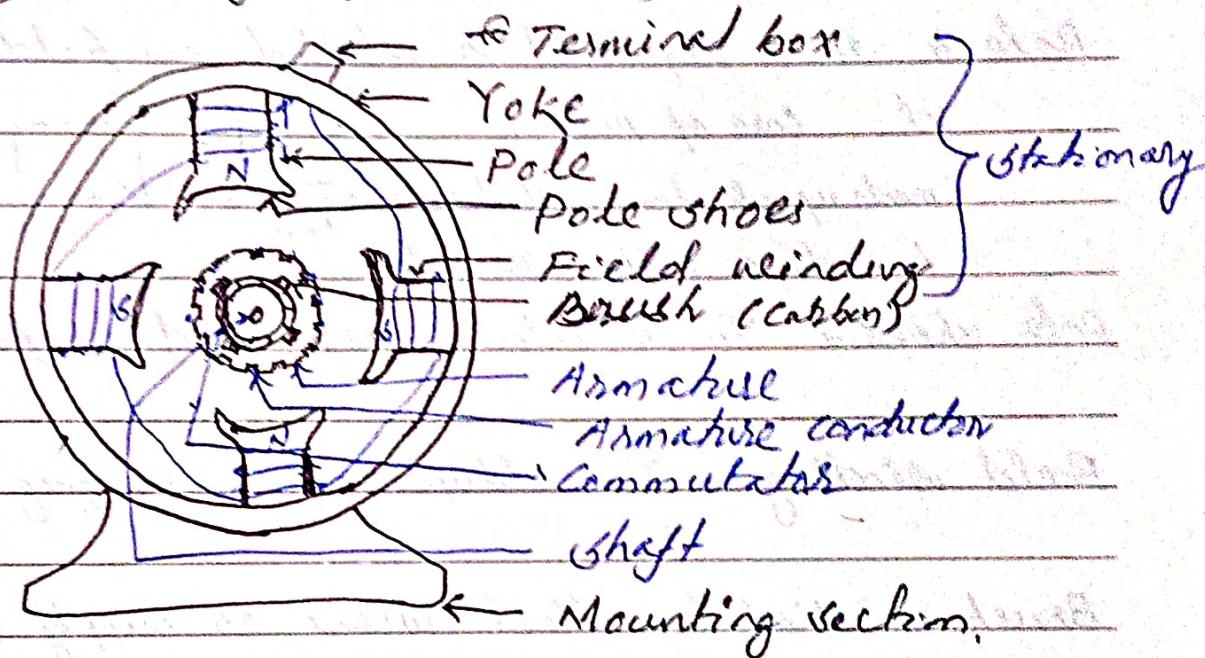
- At lower speed it can provide higher torque.
- ↓ Vice versa.

App. - Blowers, RC Drives, Elevators, Cranes.

Robotics (servo, stepper motor), Electric locomotive

DC generators - converts mechanical energy to electrical energy and it is of direct current nature used in electrolysis, electric arc, electric welding

Construction (DC gen & DC motor)



DC generator can be used as dc motor & these work on electromechanical conversion of energy.

E construction wise no change depending upon permanent magnet motor or brushless type  
construction of dc machine

Stator section (stationary) Rotor section (Rotating part)



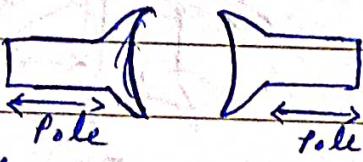
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Yoke ~ Outermost part of machine. It provides protection from external condition.

- Must be form low reluctance path (to create path for mag. flux)

Pole ~ It is required to establish mag. field. & distribute it. Core of pole is made up of laminated sheet to reduce eddy current loss.



Pole shoes ~ To distribute mag. field uniformly

Field winding ~ To buildup pole by supplying the current

Brush ~ Used to collect current or supply current to armature wire. It is made up of carbon or graphite

Armature core ~ The system or portion on which armature winding is done. It is laminated

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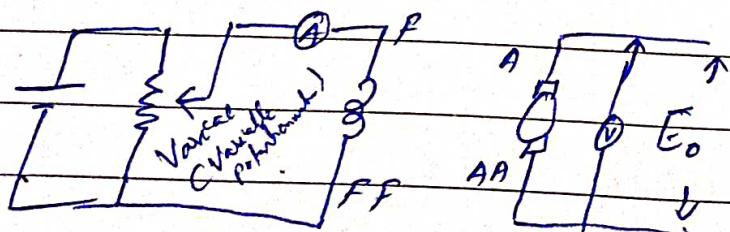
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## Characteristics of DC Generator.

- Separately excited  
Exited
- 1) open circuit characteristics (OCC)  $E_o \propto f/N = \text{const.}$   
No load characteristic or magnetisation characteristics
  - 2) load characteristics  $V_f \propto f / N = \text{const.}$
  - 3) external characteristics  $V_f \propto I_a / N = \text{const.}$
  - 4) internal characteristics  $E_f \propto I_a / N = \text{const.}$
  - 5) Regulation characteristics  $E_f \propto f_i / N = \text{const.}$

→ First two characteristics field current is varied  
other remaining / keeping const. so this can be done  
in case of Separately excited dc generator.  
And other is in case of self excited.

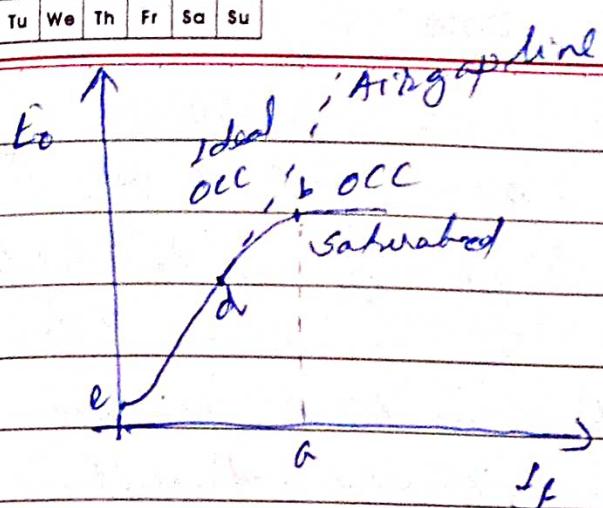
### ① OCC





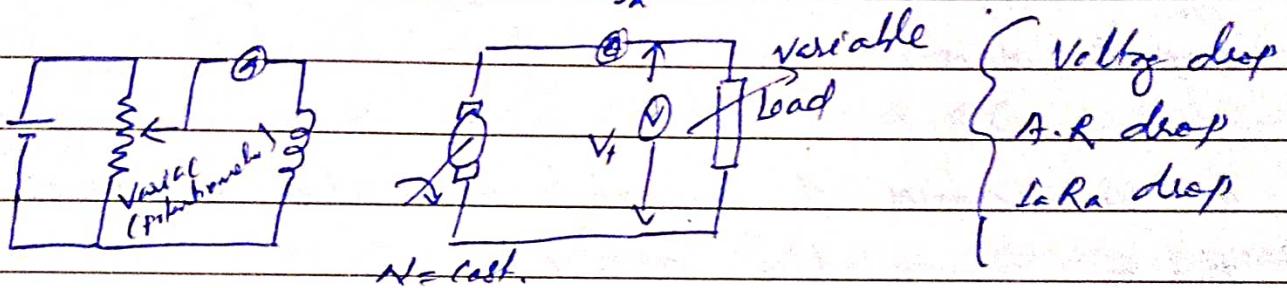
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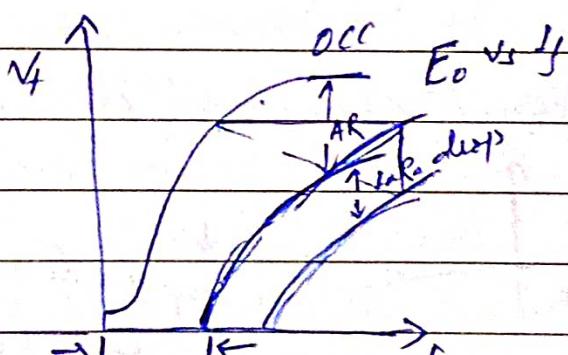


Due to the residual magnetism even though the field current is zero some emf will be induced.

② Load characteristics.  $I_a = \text{const.}$



$\delta f$	$V_t$
1	
2	
3	
4	



As  $I_a$  is const. so armature reaction demagnetizing effect of A.R. is const. So the OCC curve need shift downwards. This shift is due to demagnetization.

→ the value of  $I_f$  required to overcome the demagnetizing effect of armature reaction.

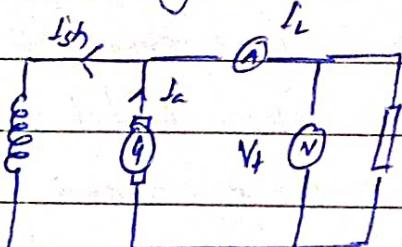
There is one more  $I_a R_a$  drop so it drops further downward and it is const. as  $I_a$  is const.



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### (a) shunt Generator



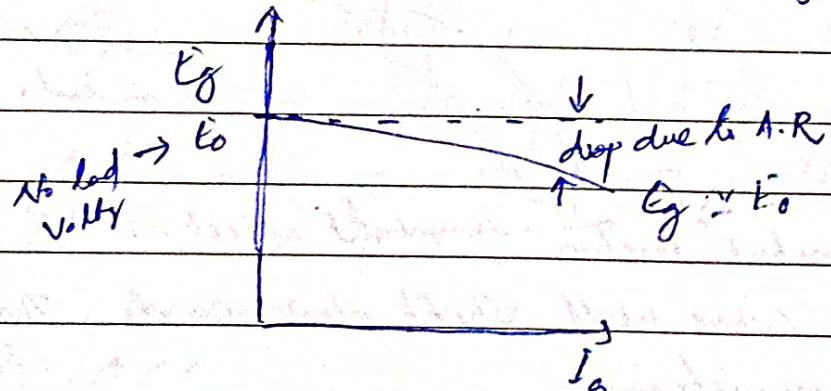
different currents are  $I_{sh}$ ,  $I_a$ ,  $I_L$

$$I_{sh} = \frac{V}{R_{sh}}, I_a = I_L + I_{sh} \approx I_L$$

Def. drop is due to armature reaction. (a) due to 1st & 2nd effect automatically drop in terminal voltage happens. When this happens  $I_{sh}$  decreases (as  $I_{sh} = \frac{V_t}{R_{sh}}$ ) or further extend drop to decrease in  $I_{sh}$  or  $I_a$ .

As  $V_t$  decreases the generated emf also decreases this is cumulative action.

### (b) internal characteristics $E_g$ vs $I_a$ / $N_o$ vs $I_a$



\*  $E_g$  is not constant as in case of no load characteristic  
because armature reaction depends on armature current ( $I_a$ )

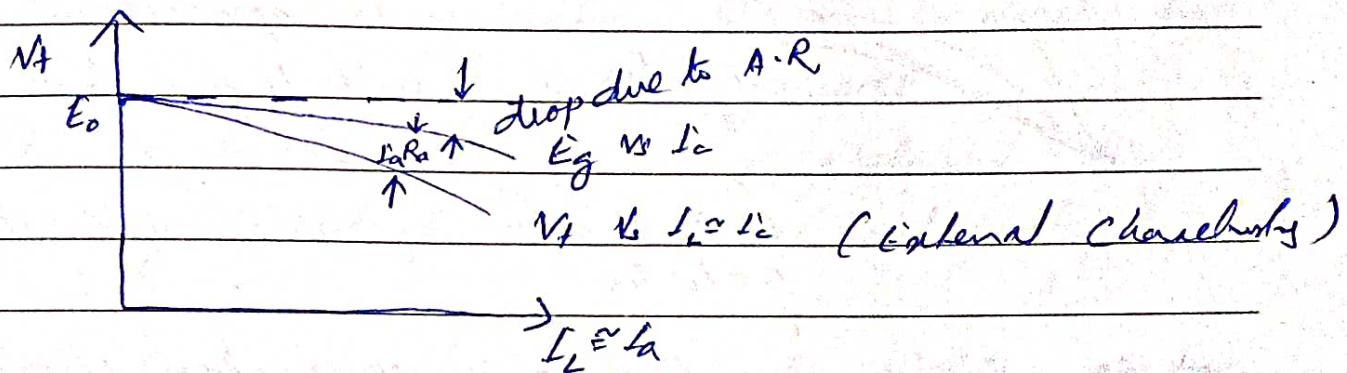
So.  $E_g$  will go on decaying with  $I_a$



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#### (4) External characteristics



So from above we can say that

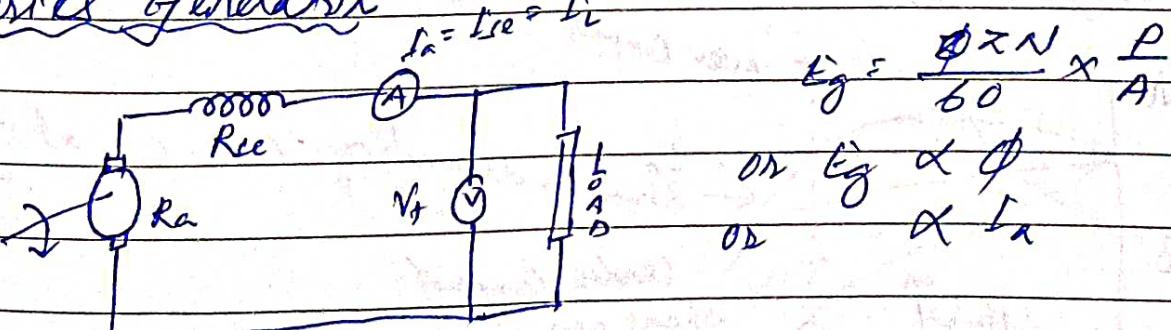
$$\text{No load } E_g = V_t + I_a R_a$$

$$\text{V.L.L } \rightarrow E_0 = V + I_a R_a + \text{drop due to } A \cdot R$$

- ⇒ Characteristics of shunt generator in drooping characteristic
- ⇒ Shunt generator can be considered as cont. flux & cont. voltage machine. (Voltage drops from no-load to full load is very less)

Application: used as voltage source for lighting  
as an exciter in alternator (dc field)

#### Series Generator



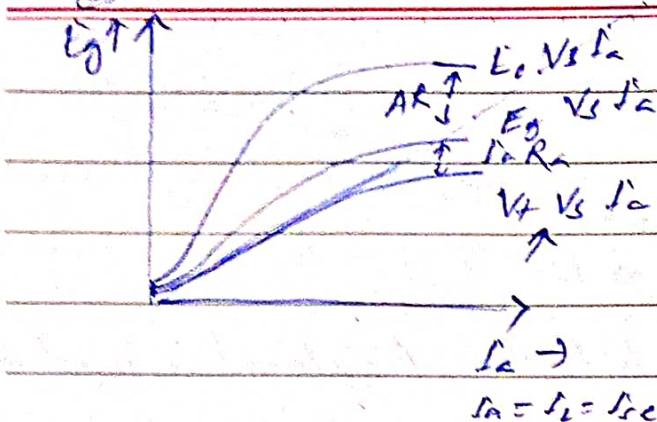


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Characteristics

$R_a + R_L$

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We can say all because under open circuit nothing will be generated as

$$I_{se} = 0$$

$$I_a = I_e = I_{se}$$

- So we can say that it is of rising characteristics
- \* As load increases the terminal voltage will increases

Application : used as voltage booster

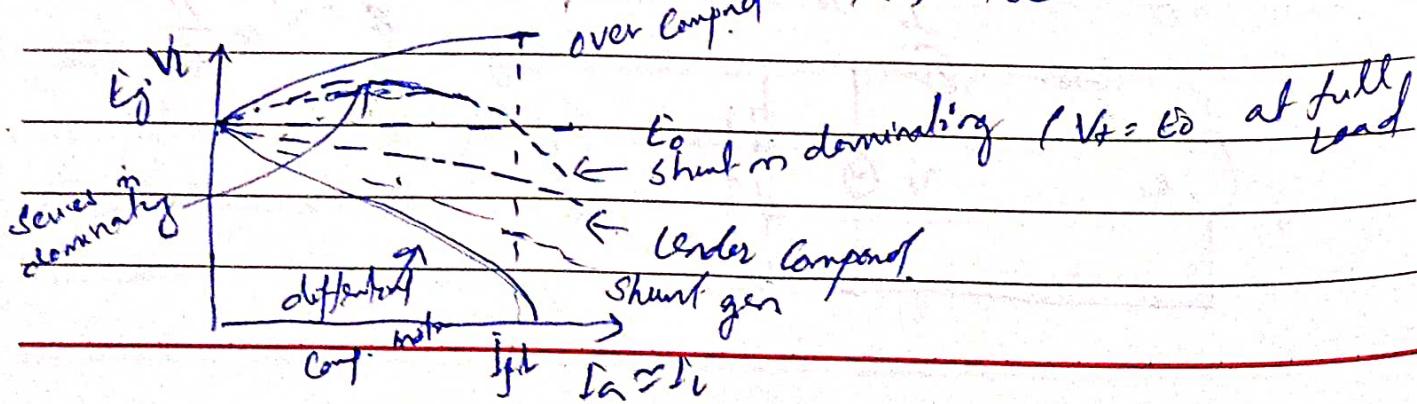
- \*  $R_a$  &  $R_L$  are also included in field circuit resistance
- \*  $R_{se} + R_a + R_L$  should be less than electrical field resistance then only we can generate some emf

Compound Generators (series + shunt field exc)

$$\text{cumulative compnd } (\phi) = \phi_{sh} + \phi_{se}$$

over excited   flat aligned   under excited

Differential " =  $\phi_{sh} - \phi_{se}$





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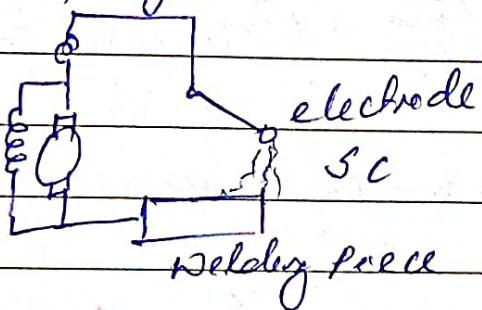
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## Applications

\* Cumulative Comp. generator can be used for voltage sources for lighting purpose.

\* Flat & level comp. gen. are preferred for exciting field of alternator. (field exciter)

\* Def. Comp. generator is used for welding purpose.



As Load  $\uparrow \rightarrow V_f$

When  $V_f$  drops automatically  
the  $E_g$  (generated emf. drops)  
hence  $\Phi$  decreases (law)

Because of this the current (shunt circ.) will be  
come down to limited range of generator & if doesn't  
damage the generator.

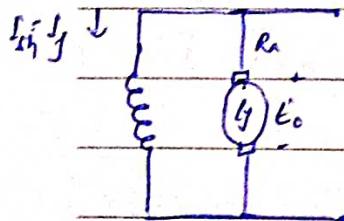


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## Voltage Build up

\* Voltage build up depends on the presence of residual flux in the pole of generator.



$E_0$  initial is due to residual flux

$$\text{e.g. } E_{0\text{res}} = k \phi_{res} \omega (\approx 1-2 \text{ volt})$$

(Shunt Gen.)

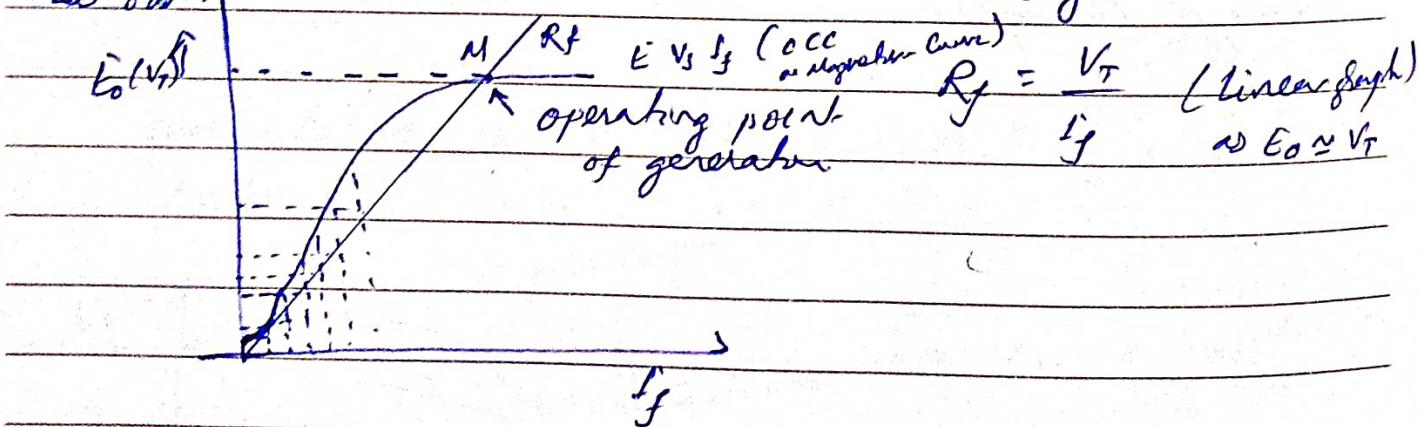
$$I_f = \frac{E_{0\text{res}}}{R_f} \rightarrow \phi \uparrow \text{ if initial } \phi \text{ was zero}$$

The flux must be greater than residual flux

so.

$$\begin{aligned} & \rightarrow E_0 \uparrow \\ & \rightarrow V_T \uparrow \\ & \rightarrow I_f = \frac{V_T}{R_f} \\ & \rightarrow \phi \uparrow \end{aligned}$$

\* flux will increase up to its saturation point if it limits the terminal voltage.



$E_0$  &  $I_f$  increases simultaneously steady state.



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Why no voltage build up in shunt gen.

(1) There may be no residual flux.

$\text{Ans} \rightarrow$  Connecting the field to an external dc current  
(It is called as flicking of field)

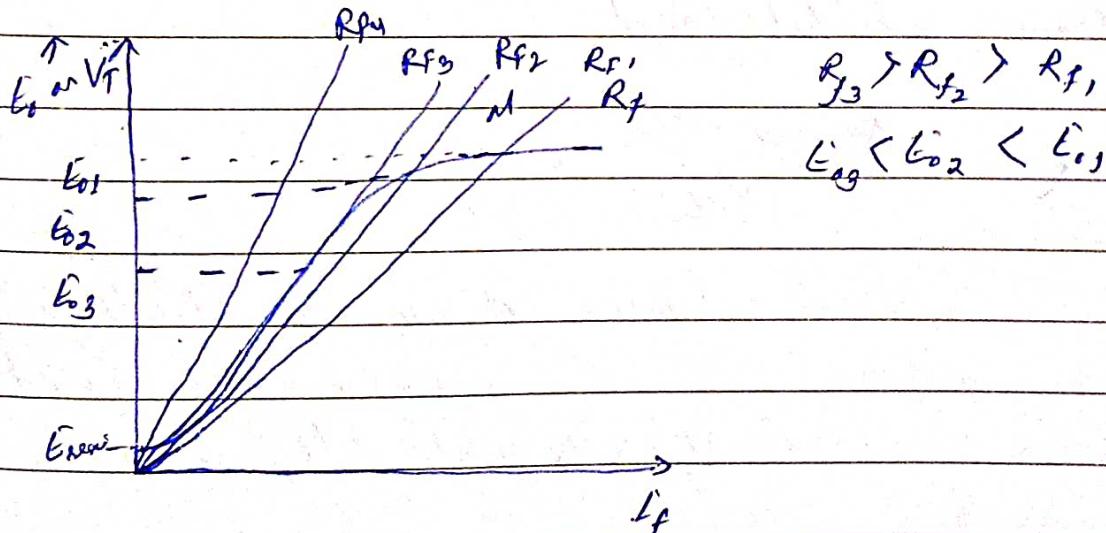
(2) The directions of rotation might be reversed. On  
 $\leftarrow \theta_{\text{max}}$  field connection could have been reversed.

$\text{Ans} \rightarrow$  either reverse the direction and check again or  
reverse the field connector.

(3) The field resistance ( $R_f$ ) might be more than the  
critical field resistance ( $R_{f, \text{critical}}$ ) [ $R_f > R_{f, \text{critical}}$ ]

$\text{Ans} \rightarrow$

Operating point is the intersection between magnetization  
curve & field resistance line ( $R_f$  line)



At  $R_{f_3}$ , the value of  $B_3$  changes significantly with  
Change in  $R_f$  or  $I_f$



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If  $R_f$  is increased beyond  $R_{f3}$  then the voltage is only residual one & no voltage build up occurs.

- \* We concluded that for any value of  $R_f > R_{f3}$  (critical field resistance) voltage will not build up because the operating point is just at residual voltage.

## Voltage Regulation

Change in the terminal voltage of a generator between full load and no load (at const. speed) is called the voltage regulation, usually expressed in % of the voltage at full load.

$$\% \text{ V.R.} = \frac{V_{NL} - V_{FL}}{V_{FL}} \times 100$$

Where  $\Delta V_{NL}$  = Terminal voltage of generator at no-load  
 $V_{FL}$  = Terminal voltage at full load.

- \* Lesser the V.R. the generator's v.r. is good and more V.R. indicates the poor voltage stability.



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⇒ Series DC generators are having poor voltage regulation i.e. V.R is more.

⇒  $V.R = \frac{V_f - V_n}{V_n} \times 100$  can be defined as the change in terminal voltage from no load to full load expressed in term of full load at const. speed.

1) Over Comp. DC generator  $V_{FL} > V_{NL}$

2) Under Comp. "  $V_{FL} < V_{NL}$

3) Level or flat "  $V_{FL} = V_{NL}$

Sequence of V.R :-

Series > Differential > over comp. > shunt > separately > (level) > flat  
↓ +ve postive -ve +ve +ve +ve zero  
V.R -ve postive (As characteristics in series i.e. V increases with load)  
V.R

## Parallel operation of DC Generators

Why or Benefits of parallel operations.

a) Efficiency is increased : DC generators perform best or very well at full load or near to full load. When load increases, two or three DC gen. are connected in parallel in order to improve efficiency ( $\eta$ )



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b) Maintenance & Repair : Maintenance is very essential for any m/c so in order to avoid interruption, the maintenance & repair of m/c can be performed.

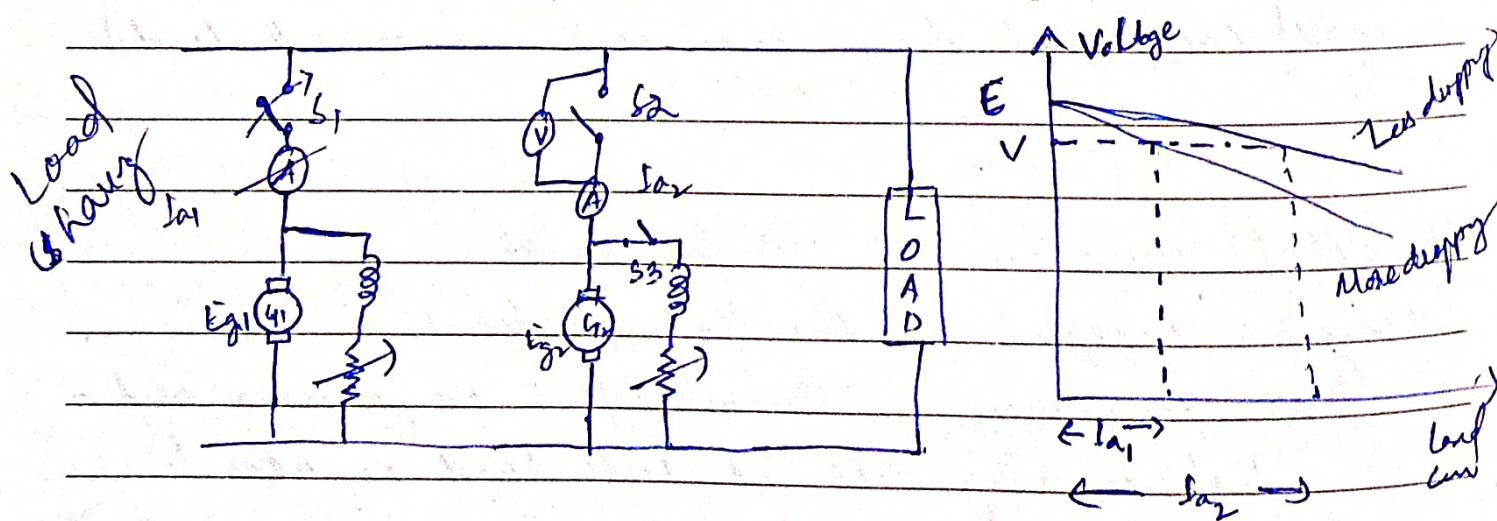
c) Reliability or continuity of Supply : Continuity of power supply can be maintained.

d) Additional demand can be fulfilled according to requirement by connecting Generators in parallel

(a)

\* Condition of parallel

Parallel operation of DC shunt Generators.



→ The generator having more drooping characteristic shall lead leading & vice versa.



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## \* Conditions for parallel operation

1) The terminal voltage of both the generators must be equal. To make equal voltage we need to vary the field current.

If voltmeter reads zero reading, both the generators are at same potential.

2) The polarity of incoming gen. must be same as the polarity of existing gen.

(3)

$$\frac{3}{G_1 T} \frac{1}{T} G_2$$

## In case of short DC gen

1) The prime mover of  $G_2$  is brought to the rated speed.

2) The field excitation is varied or adjusted to get voltage as  $G_1$ .  $T_2$  chosen by  $S_2$

3) Then close the switch  $S_2$

4) Gradually increase  $I_f$  of  $G_2$  till it supplies the required load.

$$I_2 = \frac{G - V}{R_2}$$

\* To increase the load share of any generator we have to increase the field of that particular generator.



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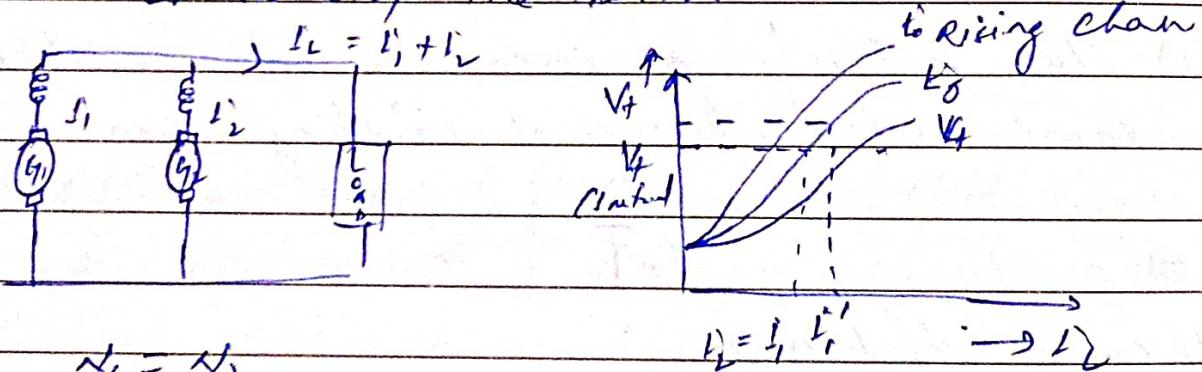
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\* Decrease field current of generator which tends to share less load.

⇒ Shunt & Comp. Generators used for power supply

### Series DC Generators

→ Used in railway for regenerative braking or to stop the train.



$$n_1 = n_2$$

$$E_g1 = E_g2$$

$$f_1 = f_2$$

Suppose the load increases &  $E_g1 > E_g2$  then (RPM)

$f_1 > f_2$  so as  $I_2$  decreases  $f_2 \downarrow$

$E_g f_1 > E_g f_2 \downarrow$  &  $E_g1 > E_g2 \downarrow$

Because of this the generator may spark or damaged



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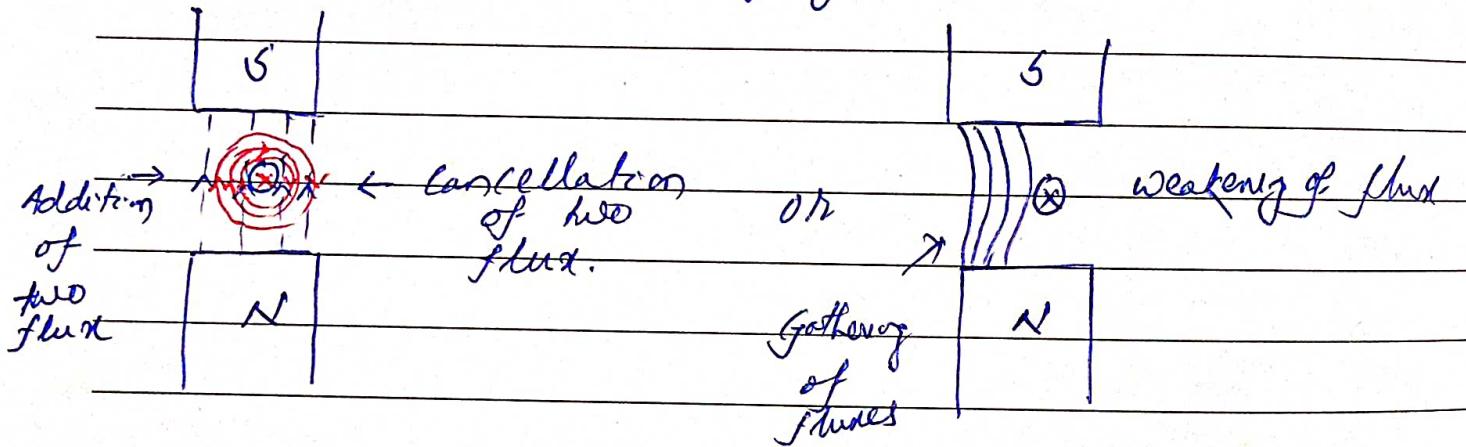
# DC MOTORS

Date / /

Working Principle :-

When a current carrying conductor is placed in a magnetic field, it experiences a mechanical force.

- flux is created by a field winding.
- Armature winding plays a role of current carrying conductor.



\* Force is experienced from high flux density to low flux density.

Direction of Rotation of Motor.

Magnitude of the force experienced by conductor (armature) is given as

$$F = BLI \text{ newton}$$

Where  $B$  = Mag. flux density produced by field winding

$L$  = Active length of conductor

$I$  = Mag. of current carried by conductor



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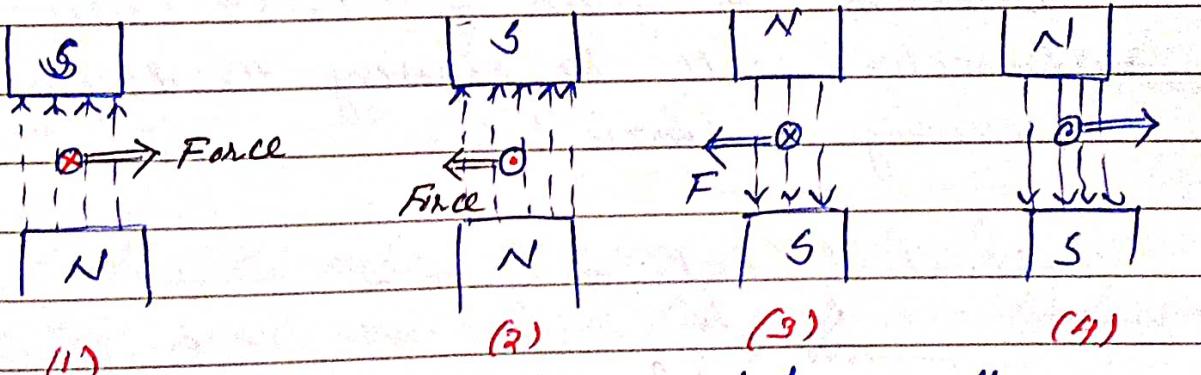
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This force is a vector quantity & its direction is determined by Fleming's Left hand Rule (LHR)

~~Fleming's left hand Rule (LHR)~~

The rule states that, if we stretch the three finger of the left hand namely the forefinger/pinky finger, middle finger & thumb such that they are mutually perpendicular to each other.

Now the pinky finger is in the direction of magnetic field and the middle finger is in the direction of current then the thumb gives the direction of the force experienced by the conductor.



To reverse the direction of rotation either of main field produced by the winding is reversed (case 1, 3) or direction of the current passing through armature is reversed (case 1, 2)



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## Back Emf ( $E_b$ )

Once motor starts rotating its conductor will cut the magnetic flux produced by field winding. So by Faraday's law of electromagnetic induction emf is induced in the armature coil of dc motor just like in case of ac dc generator and called as back emf.

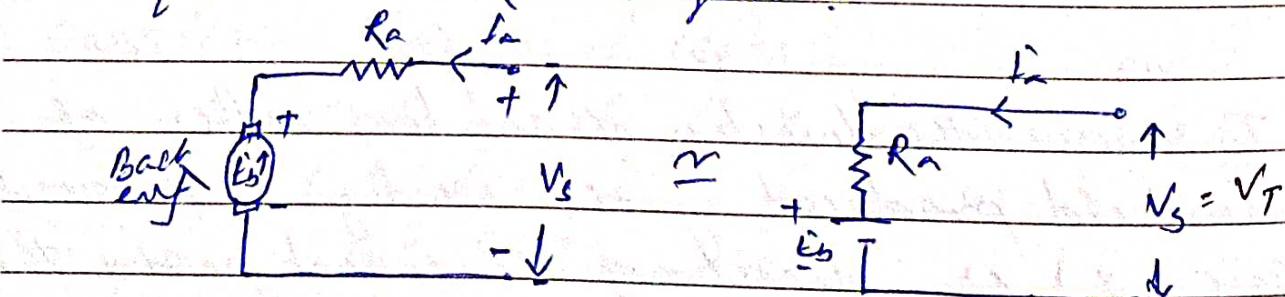
\* Back emf is generated by the generating action (moving conductors cutting the magnetic flux)

$$\text{ie } E_b = \frac{\text{ZPN}}{60A} \quad \text{Emf eqn of motor.}$$

Direction is given by Lenz's Law.

$\Rightarrow$  Direction will be opposing the supply voltage according to Lenz's Law.

Equivalent ckt. diagram :



Voltage Eq?

$$V_s = I_a R_a + E_b + V_{brush}$$

$V_s$  = Supply voltage       $I_a$  = Armature current

$R_a$  = Armature resistance       $E_b$  = Back emf       $V_{brush}$  = Brush drop



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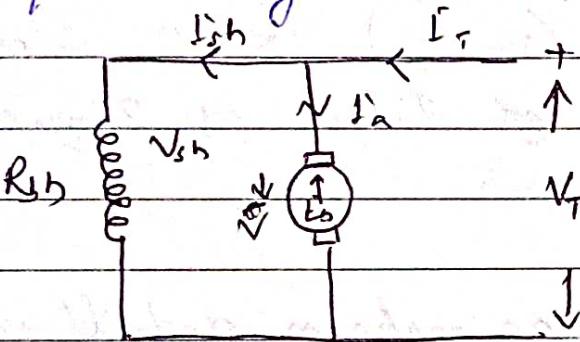
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$$I_a = \frac{V_T - E_b}{R_a}$$

$V_T$  &  $R_a$  are constant. ( $E_b \propto V_T$ )

Significance of back emf.

- \* DC m/c is a self regulating machine because of back emf.
- \* It enables the dc motor to draw as much armature current as is sufficient to develop the torque required by the load.



$V_T$  = Supply voltage.

$R_{sh}$  = shunt field resistance

$I_{sh}$  = shunt current

(DC shunt motor)

Now

$$I_a = \frac{V_T - E_b}{R_a} \quad \text{or} \quad I_a \propto E_b \quad \left\{ \text{As } V_T \text{ & } R_a \text{ are const / fixed} \right\}$$

Case 1 (When DC motor is running on no-load)

- \* The torque required is less. [Torque depends upon  $I_a$ ]

$$I_a = \frac{V_T - E_b}{R_a} \uparrow \text{(1st)}$$

As m/c is running at no load the machine starts rotating freely or at high speed so  $E_b$  increases. Hence  $(V_T - E_b)$  decreases so  $I_a$  decreases. Finally torque decreases as adjusted to required torque.



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Case-2 (When a dc motor is suddenly loaded)

\* Torque required to drive load increases.

If load increases then speed of rotation decreases then  $E_s$  reduces

$$i_a = \frac{V_f - E_s}{R_a} \downarrow (i_a)$$



This results decrease in  $(V_f - E_s)$  &  $i_a$  increases and results in increase of Torque as Torque depends on  $i_a$ .

Case-3 When the load on dc motor is gradually reduced  
[ Right load to Torque required will decreases gradually.

$$i_a = \frac{V_f - E_s}{R_a}$$

1) Speed gradually decreases as load decreases gradually  
 $\therefore E_s \uparrow$

As  $E_s \uparrow$  the term  $(V_f - E_s)$  decreases &  $i_a$  decreases and results gradual decrease in torque.



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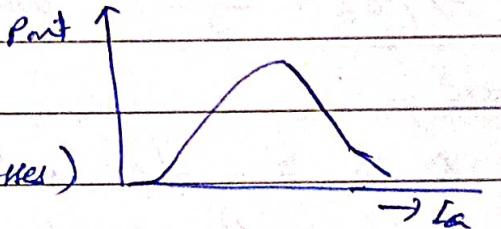
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Condition for max power output in Generator developed in DC Machine

\* The dc generator will deliver maximum power when the terminal voltage ( $V_T$ ) is half of the emf.  $\left\{ \begin{array}{l} I_a = \frac{E_a}{2R_a} \\ N_g = E_a - I_a R_a \\ = E_a - \frac{E_a}{2R_a} R_a \\ \frac{V_T}{E_a} = \frac{1}{2} \end{array} \right.$

$$* N = \frac{P_{out}}{P_{in}} = \frac{V_T I_a}{E_a I_a} = \frac{V_T}{E_a} = \frac{1}{2}$$

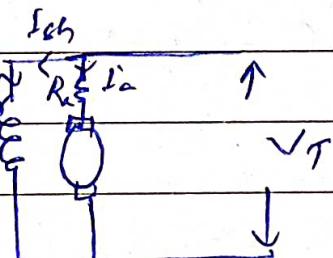
$$= \frac{1}{2} = 50\%$$



\* If all conditions are taken (losses) the  $\eta$  will be less than 50%.

### DC Motor

$$E_b = V_T - I_a R_a \quad \text{--- (1)}$$



$$P_{de} = E_b I_a \quad \text{--- (2)}$$

$$\text{ie } E_b I_a = V_T I_a - I_a^2 R_a \quad \text{--- (3)}$$

For maximum power delivered,  $\frac{d}{dI_a}(P_{de}) = 0$

$$\Rightarrow \frac{d}{dI_a}(P_{de}) = \frac{d}{dI_a}(V_T I_a - I_a^2 R_a)$$

$$\Rightarrow 0 = V_T - 2I_a R_a \quad \text{--- (4)}$$

$$\Rightarrow \boxed{I_a = \frac{V_T}{2R_a}} \quad \text{--- (4)}$$



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Putting the value of  $I_a$  from eq(7) to in eq(0)

$$I_b = V_T - \frac{V_T}{2R_a} R_a =$$

$$\Rightarrow E_b = V_T - \frac{V_T}{2}$$

$$\Rightarrow \underline{E_b = \frac{V_T}{2}} \quad \text{or} \quad E = \frac{V}{2}$$

\* In a dc motor power developed is maximum when the Back EMF ( $E_b$ ) is half of the supply terminal voltage.

\*  $n < 50 \times$

\*  $I_a = \frac{V_T}{2R_a}$

Torque Equation of DC motor. :-

$\Rightarrow$  The turning or twisting force acting on a body about an axis is termed as Torque ( $T_o$ )

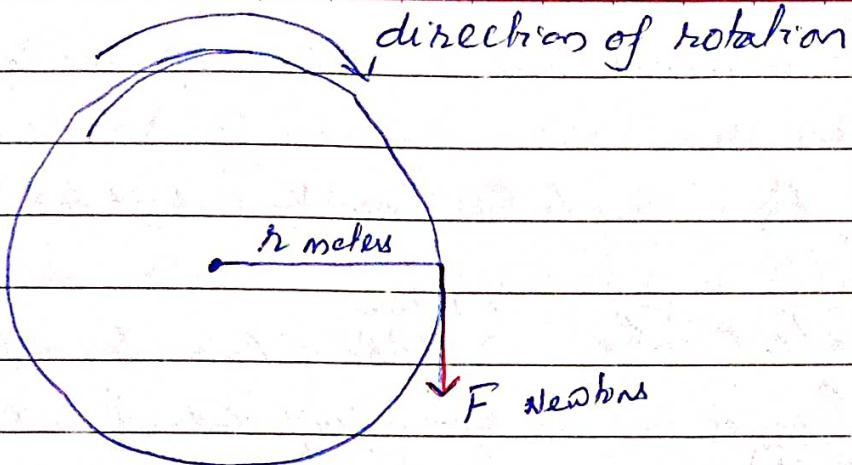
or

$\Rightarrow$  Torque is a quantitative measure of tendency of a force to cause rotational motion in a machine



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Let us assume that a motor is running in clockwise direction as shown above with an angular velocity of  $\omega$  rad/sec

$$\text{where } \omega = 2\pi f$$

Now we can say

$$D = \frac{2\pi \cdot N}{60}$$

$$\left. \begin{array}{l} \text{where} \\ f = \text{frequency} \quad f = \frac{1}{T} \end{array} \right\}$$

$$N = \text{rpm} : 60 \text{ sec} \rightarrow N \text{ rev/min}$$

$$\text{Time for 1 revolution} = \frac{60}{N}$$

$$\therefore T = \frac{60}{N}$$

$$\therefore f = \frac{N}{60}$$

We know that the work produced in a given machine can be given as  $T = F \times r$  — ~~<sup>if</sup>~~ <sup>Nm</sup> ~~force x displacement~~  
work done in one revolution  
 or  $W = \text{Force} \times \text{distance covered}$ .

$$W = F \times 2\pi r \quad (\text{Circumference})$$

The mechanical power developed by the motor is given by

$$P_{\text{out}} = \frac{\text{work done}}{\text{time}} = \frac{F \times 2\pi r}{60/N}$$

$$= F \cdot r \times \frac{2\pi N}{60} \text{ or } T_a \times W$$

$$\boxed{P_{\text{out}} = T_a \times \frac{2\pi N}{60}} - \text{Ans} \quad (4)$$



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Now the electrical input power of DC machine can be given as

$$\text{We know } E_b = V_f - I_a R_a \text{ or } V_f = E_b + I_a R_a$$

$$\Rightarrow V_f I_a = E_b I_a + I_a^2 R_a \quad \left\{ \begin{array}{l} \text{Where } \\ I_a^2 R_a = \text{Power loss in form of heat} \\ \uparrow \text{power in armature} \quad (\text{Neglected as ideal case}) \end{array} \right.$$

$$P_{in} = E_b I_a - \textcircled{A}$$

$$\left\{ \begin{array}{l} \text{on } P = VI \quad \& \text{Voltage} \\ \text{is armature in } E_b \text{ & } I_a \end{array} \right.$$

$$\therefore E_b \textcircled{A} = E_b (B) \quad \left\{ \begin{array}{l} \text{is armature in } E_b \text{ & } I_a \\ \text{Power in armature} = \text{gross armature torque} \times \omega \end{array} \right.$$

$$E_b I_a = I_a \times \frac{2\pi N}{60}$$

$$\Rightarrow I_a = \frac{E_b I_a}{2\pi N} \times 60 = \frac{\frac{z\phi NP}{60A} \times I_a}{2\pi N} \times 60$$

$$\Rightarrow I_a = \frac{z\phi NP \times 60}{60A \times 2\pi N} \times I_a = \frac{\phi z P I_a}{2\pi A} \text{ A.m}$$

$$\Rightarrow \boxed{I_a \propto \phi I_a} \quad \text{As } z, P, A \text{ & } 2\pi \text{ are const.}$$

$$\Rightarrow \boxed{I_a = \frac{0.159 \times \phi z P I_a}{A}} \quad \text{W.M. Armature torque developed in a.c. motor.}$$



\* Practically the mechanical power developed in the armature is transmitted to the load through the shaft of the motor.



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- \* It is impossible to transmit the entire power developed by the armature to the load because while transmitting the power through shaft, the power losses occurs due to friction, windage and iron core or iron loss.
- \* The torque required to overcome this loss is called as loss torque ( $T_L$ ) or called as stray loss of the motor.

⇒ The torque available at the shaft for doing the useful work is known as load torque or shaft torque & denoted as  $T_{sh}$  or  $T_h$

$$\therefore T_a = T_L + T_{sh} \text{ or } T_a = T_L + T_{sh}$$

Hence  $T_{sh} \times \text{Angular speed}(\omega) = \frac{\text{Power available at shaft of motor.}}{\text{Output}}$

$$\text{or } T_{sh} = \frac{\text{output}}{\omega}$$

$$\Rightarrow T_{sh} = \frac{\text{output}}{2\pi N/60}$$

$$\Rightarrow T_{sh} = 9.55 \times \frac{\text{output}}{N}$$



1)  $N$  vs.  $I_a$  } electrical characteristics

2)  $T$  vs.  $I_a$  }

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3)  $N$  vs.  $T$  }

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## Characteristics of DC Series Motor.

(a) Speed ~ Armature Current & N.R.E.M.)

$$\text{As we know } E_b = \frac{2\pi N}{60} \times \frac{\phi}{\theta}$$

\*  $I_a R_a$  drop is neglected. So  $E_b$  is const.

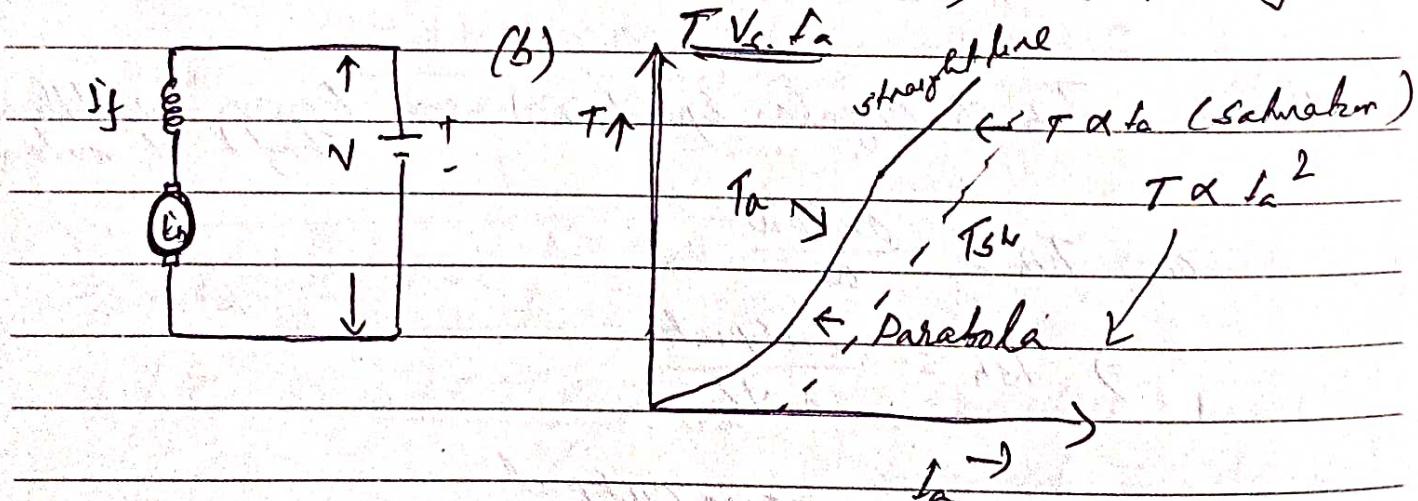
$$\text{so } N \propto \frac{E_b}{\phi} \therefore N \propto \frac{I_a}{\phi} - ①$$

$$\text{Again } E_b = V - I_a R_a$$

$$\text{Power}_{op} = \frac{2\pi N T}{60} \quad E_b \propto \phi \rightarrow I_a$$

$$\Rightarrow T = 0.159 \phi^2 I_a \frac{\rho}{A} - ②$$

$$\Rightarrow T \propto \phi I_a - (2) \quad \text{if } \phi \propto f$$



$$\text{As } T_b = T_a - \text{loss.}$$



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(B) Speed vs t

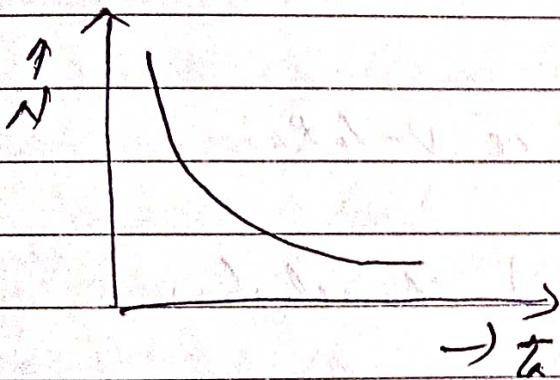
(C) Speed Vs Torque

As we know  $N \propto \frac{1}{T_a}$  — 1

and  $T \propto I_a^2$  — 2

∴ from A & B) we can write

$$N \propto \frac{1}{I_a^2}$$



Observation

→ When <sup>torque</sup> speed is less speed is extremely high. Because of this reason the series motor shouldn't be operated under no load.

$$\Rightarrow T \propto I_a^2 : N \propto \frac{1}{I_a^2} \Leftrightarrow P \propto NI^2$$

Amount of extra power required to start with high torque is very less i.e. size in power is less.

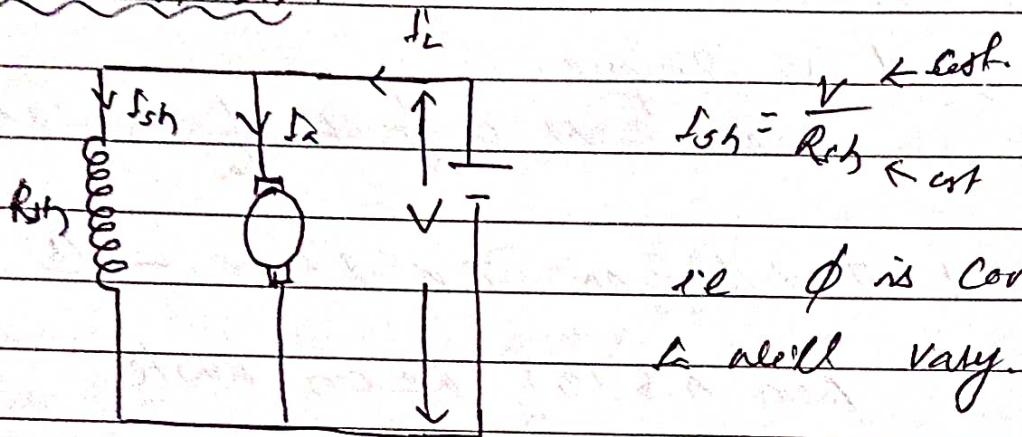
Application is High starting torque (Locomotive or train)  
Hoist, electric trains, cranes etc.



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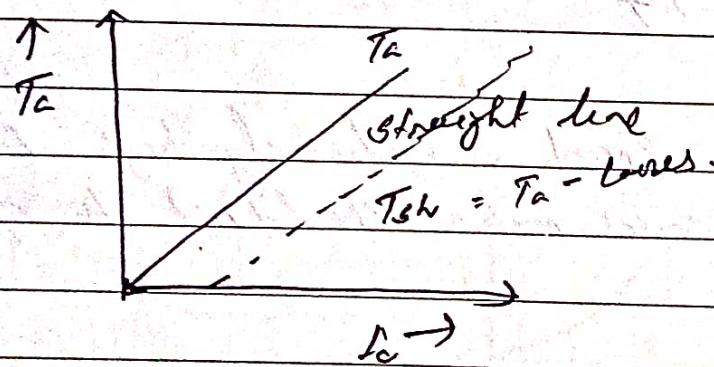
## Shunt motor.



$$T_a \propto I_a^2 \Rightarrow T_a \propto I_a$$

$$N \propto \frac{E_s}{\phi} \Rightarrow N \propto E_s \text{ i.e } V - I_a R_a$$

### a) Torque ( $T_a$ ) & Armature ( $I_a$ ) characteristics

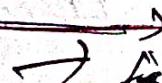


### b) Speed (N) Vs I\_a

$$N \propto E_s \text{ or } V - I_a R_a$$

→ 5 to 15%  
from N�el  
to full load

✓ Const speed motor





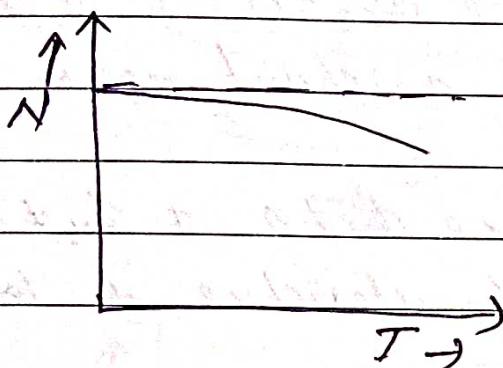
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### c) Torque ( $T$ ) Vs Speed ( $N$ )

As  $T_d \propto \omega$  so we can replace  $T_d$  by  $T_a$



Observation is

- 1) Speed regulation is very less. / mc can operate at no-load condition.

Machine tools, lathe machine, wood working mc reciprocating fans. { const speed (approximately) can be used upto 1.5 FL blower & fan { geo-lath mc, centrifugal pumps, machine tools

Disadvantages is  $T_d \propto I_c$  (series motor produces more torque)

### Compound motor

$$\rightarrow \text{Cumulative} = \phi_{se} + \phi_{sh}$$

$$\rightarrow \text{Differential} = \phi_{sh} - \phi_{se}$$

Practically only cumulative motor is used because it has the characteristics of both series & shunt motor.



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## (e) Cumulative Motor ( $\Phi_s + \Phi_a$ )

\* At no-load  $I_a = 0$  i.e.  $\Phi_a = 0$  &  $\Phi_s \neq 0$

∴ works as a shunt motor.

\* ~~For~~ full load  $I_a \neq 0$  i.e.  $\Phi_a \neq 0$  &  $\Phi_s \neq 0$

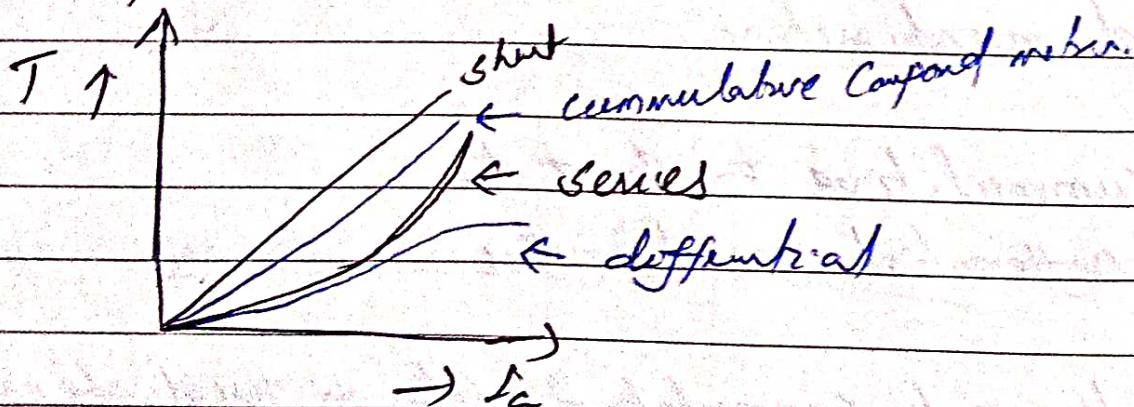
It combines the characteristics of both series & shunt motor.

1) As load increases then the series field will starts dominating the ~~field~~ shunt field as  $\Phi_a \propto I_a$

### Applications :-

Where there is enough fluctuation in the load.

Used in Electric shovels, Metal-stamper, compressor etc.

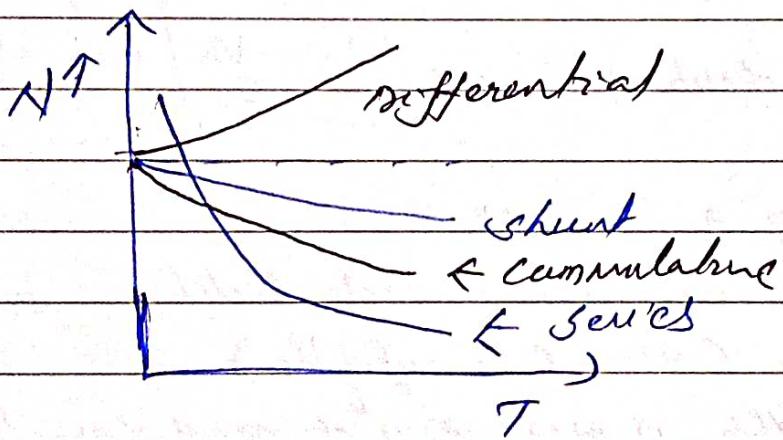
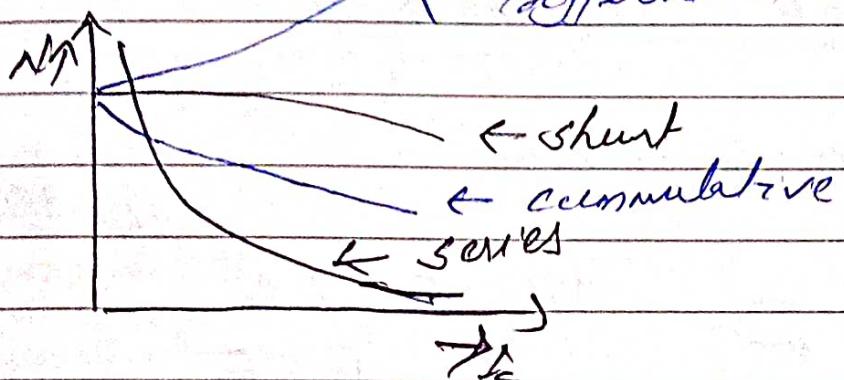




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$\rightarrow$  Differential C. motor.



use

Differential C. motor is used for experimental purpose in laboratory to get constant speed drive.

Shunt

Approximately Const. Speed  
[Const 1.5 times of FL]  
used upto

Lathe m/c, centrifugal pump, machine tools, Blowers Fan, Reciprocating pumps, Trolley

Gears

Variable speed,  
High starting torque

Tractor, Trolleys  
Cars, Cranes, hoist etc  
conveyer.

Compound  
(cumulative)

Variable speed

High starting torque  
(intermittent load)

Intermittently where

high starting torque is  
used to operate at reduced  
speeds.

e.g. shears, elevators, lifting

punching m/c, Shears,

air compressor, mills, Ice machine, printing press

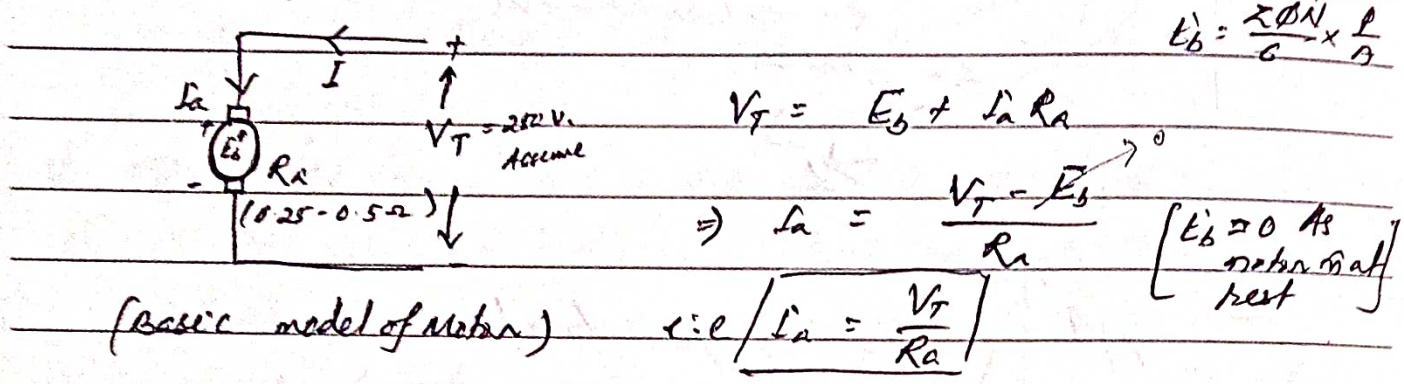


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## Starting methods of DC motor is

Need of starter in DC motor



$$I_a = \frac{250}{0.05} = 100 \text{ A OR}$$

When motor is started from rest condition is such case  $I_a \approx V_T$  (where  $R_a$  is negligible)

Because of this 1) N/C may be damaged. (carbonine windy)

2) Voltage drop in system (As it draws huge current)

To avoid all these <sup>unwanted</sup> conditions we need starter where we add external resistance in series with the armature resistance.

Starter is a) 3-point starter.  $\rightarrow$  shunt motor

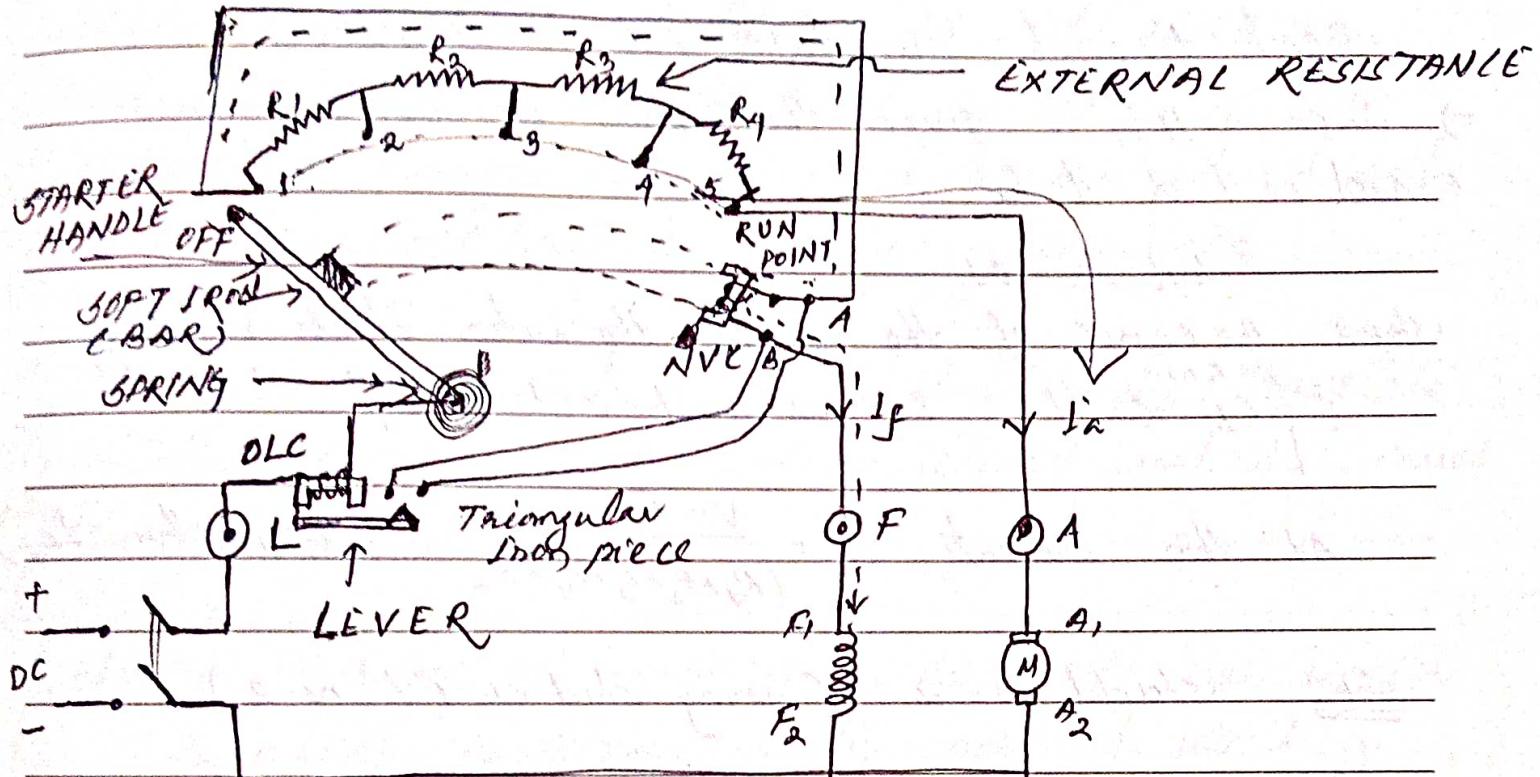
b) 4-point starter.  $\rightarrow$  compound motor.

2-point & series motor



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### (3- POINT STARTER )

OLC = over load coil

NVC = No voltage coil.

field circuit

\* R<sub>NVC</sub> is connected in series with field resistance.

As value of R<sub>NVC</sub> is very small it doesn't affect field flux.  $R_{NVC} \ll R_f \Rightarrow \phi = \text{constant}$

Armature circuit-

step)  $I_{a_0} = \frac{V}{R_f + R_2 + R_3 + R_g + R_a}$

Because of this extra resistance ( $R_f + R_2 + R_3 + R_g$ ) the



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current through armature can be limited.

$\Rightarrow$  once supply is given the motor produces torque.  
~~constant i.e.  $T \propto \Phi I_a$~~

Step no - 01

Step 2 Because of the torque the motor starts rotating

$E_f$  <sup>we know</sup>  $\propto N$  so as  $(Speed \propto N) \uparrow \rightarrow E_f \uparrow$

At this instant  $I_a' = \frac{V - E_b}{(R_1 + R_2 + R_3 + R_d) + R_a}$  ( $I_a'$  is less than  $I_{a_0}$ )

Step 3 Shaded no - 2 (moving shaded no 1 to no - 2)

{ Torque initially is increased with the help of field flux as  $T \propto \Phi I_a$ . Once torque is established the motor starts accelerating & speed ( $N$ ) increases.

Because of this  $E_f$  increases ( $E_f \propto N$ ) & results in reduction of armature current ( $I_a$ )

$$I_a = \frac{V - E_b}{R_{ext} + R_a}$$

To increase the armature current we move the shaded to next point to cut off the external resistance in phase wise manner.



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Once it reaches shield no-05, the NVC is carrying the current to the field circuit. As current is passing through it, it will be magnetized and hold the soft iron bar from going back.

Spring is provided for controlling purpose. Once the power goes the NVC is demagnetized and bar comes to original position.

NVC - Purpose of NVC is to hold the bar under running conditions under normal working conditions.

If for some reason the current is zero / Power goes then automatically the current passing through NVC is zero & then bar would return to original position.

OLC ~ It is connected in series with armature & field coil. If total or line ~~not~~ current is passing through OLC.

Under normal operating conditions it will operate as usual but whenever load crosses some limit (permissible) of motor. It will get magnetized.

Once magnetized the bar of lever will be attracted & triangular iron piece will



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short circuit the two terminals A & B. Because of this the current ~~will~~ through NVC will be bypassed & demagnetized and <sup>will</sup> release the bar and handle will come to off position.

### Field Current

Once starter handle moves forward the extra resistance will be added to field circuit.

$$f_f = \frac{V}{R_f + R_{NVC} + R_{External}}$$

\* As  $R_{NVC} + R_{External}$  is very less compared to  $R_f$  so current will be not affected.

OR Initially at shd no. 1 the flux produced is more and finally shd no. 05 the flux produced is equal to rated value of flux. (Tx Ø<sub>ra</sub>)

### Disadvantages (above rated speed)

- 1) Field control of motor can't be done as if V<sub>f</sub> and NVC is demagnetized it may trip.

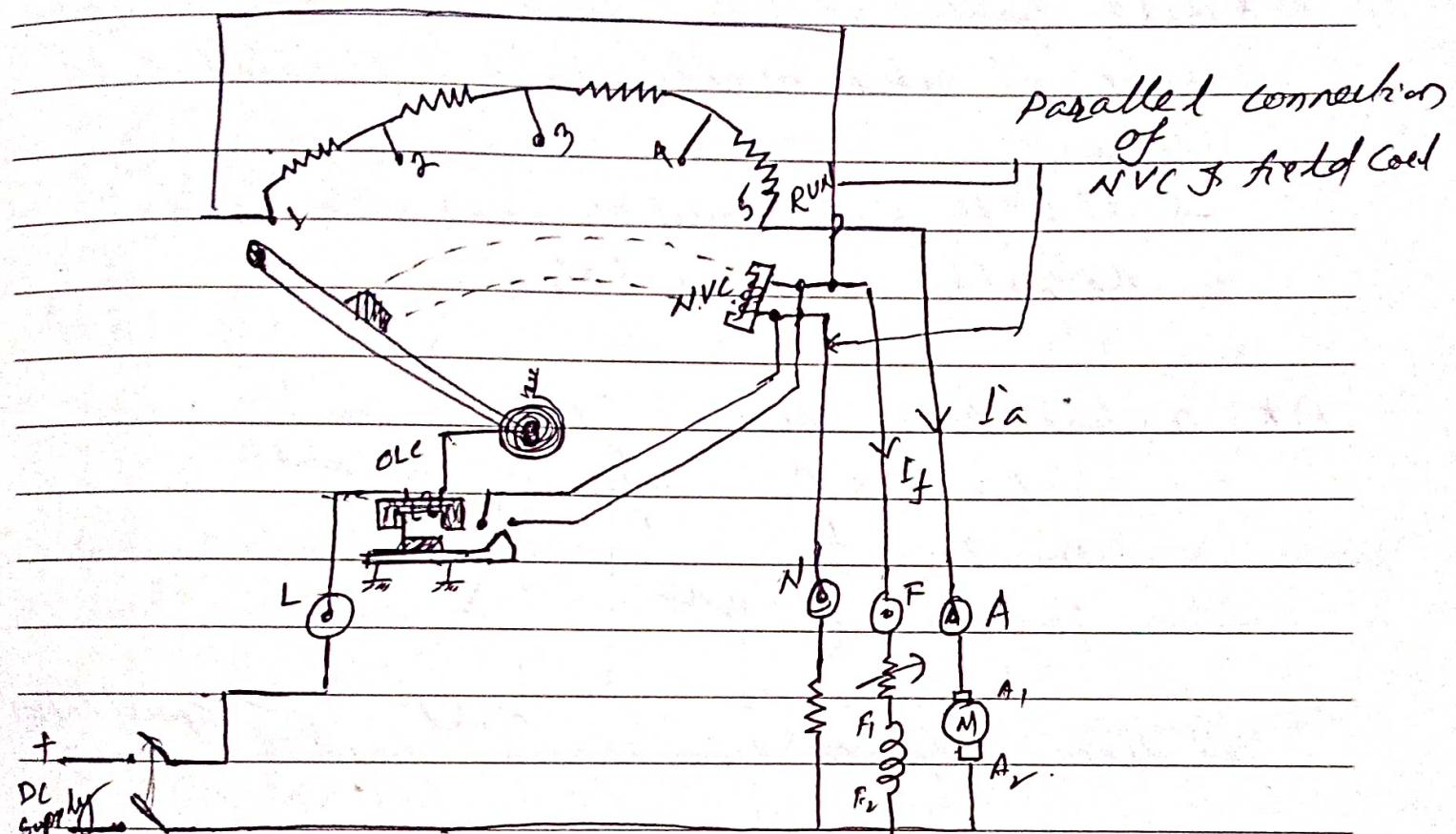
To overcome this problem we go for Four point starter. (To make NVC independent of f<sub>f</sub>)



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## 4- Point starter.



(4- Point starter.)



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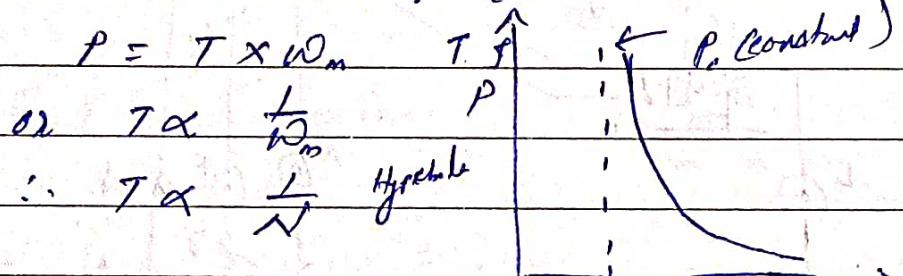
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## Speed Control of DC Shunt Motor

a) \* Base speed is It is the speed at which the motor runs at rated armature voltage & rated field current.

→ If it is the rated speed or name plate speed of the motor (Shaded / Name)

b) \* Constant Power Drive or (Const. Power)

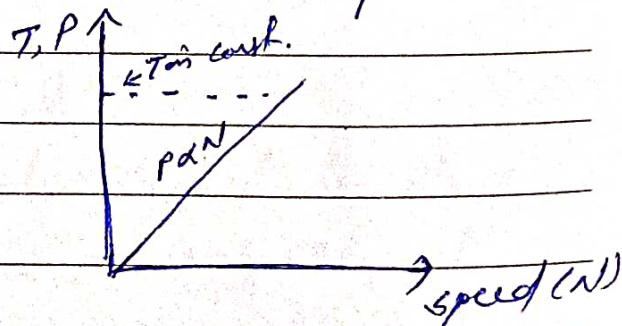


The drive in which power remains constant.

c) Constant Torque drive (Drive in which Torque is const.)

$$\text{power} \rightarrow P = T \omega_m \quad \downarrow \omega_m$$

$$\rightarrow P \propto T \quad \text{or} \quad P \propto N$$



Speed can be controlled by

✓ Field flux control method / Field Weakening / Flux Weakening / field generation method.



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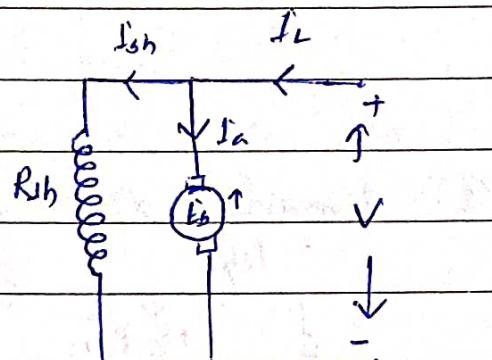
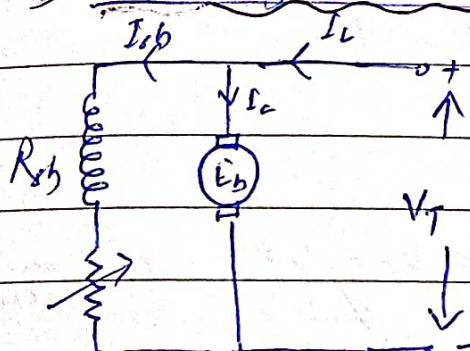
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2) Armature resistance control method.

~~3) Armature voltage control method.~~

1) Field flux control method

(15)



Externally  $R_f$  is connected in series  
with field winding

DC shunt motor under  
Normal conditions

$$\text{Eqn} \quad I_{sh} = \frac{V}{(R_{sh} + R_f)}$$

$$\text{Eqn} \quad I_{sh} = \frac{V}{R_{sh}} \quad \text{Normal condition}$$

As  $I_{sh} \downarrow$   $\phi \downarrow$

So we know  $N \propto \frac{\epsilon_b}{\phi} \propto \frac{V_f - I_a R_a}{\phi}$

Then speed will increase with decrease of flux  $\phi$   
and speed increases above rated speed (Unrated)

flux can't be increased above rated value.

Features of method:

\*  $N$  (Speed) above rated speed can be obtained

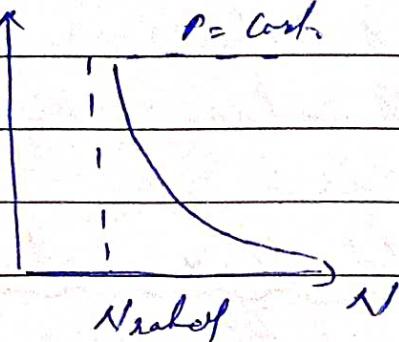


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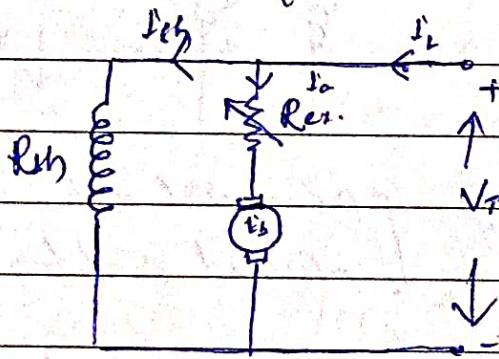
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\* Constant Power drive  $T_p \uparrow$   $P = \text{const.}$

\* Variable Torque drive.



(2) Armature resistance method.



$$\text{i) } N \propto \frac{V_f - I_a R_a}{\phi} \quad [\text{Normal Condition}]$$

$$\text{ii) } N \propto \frac{V_f - I_a (R_a + R_{ext})}{\phi}$$

$$\text{iii) } E_b = V_f - I_a (R_a + R_{ext})^{\pi}$$

$$\text{iv) } I_a = \frac{V_f - E_b}{R_a} \quad [\text{Normal}] \quad I_a = \frac{(V_f - E_b)^{\pi}}{R_a + R_{ext} \pi}$$

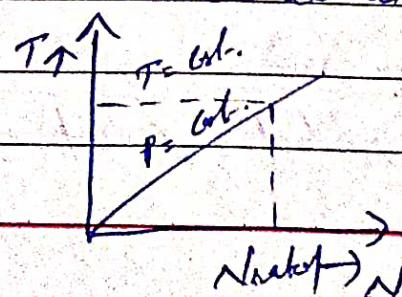
i.e.  $I_a$  is constant &  $\phi = \text{const.}$  (for direct m/s)

Feature.

As we know  $T \propto \phi I_a$  or  $T \propto I_a$

∴ Torque is almost const. in this method.

② Speed is below rated speed as  $E_b \downarrow$  with external resistance as in armature.



③ Variable Power drive



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(2) Voltage drop occurs due to external load.  
Thus  $I^2R$  (power loss) occurs. & unnecessary heating of machine occurs.

(3) Armature voltage control method.

(Not applied to small DC practically)

→ It is applicable to large method because it is an expensive.

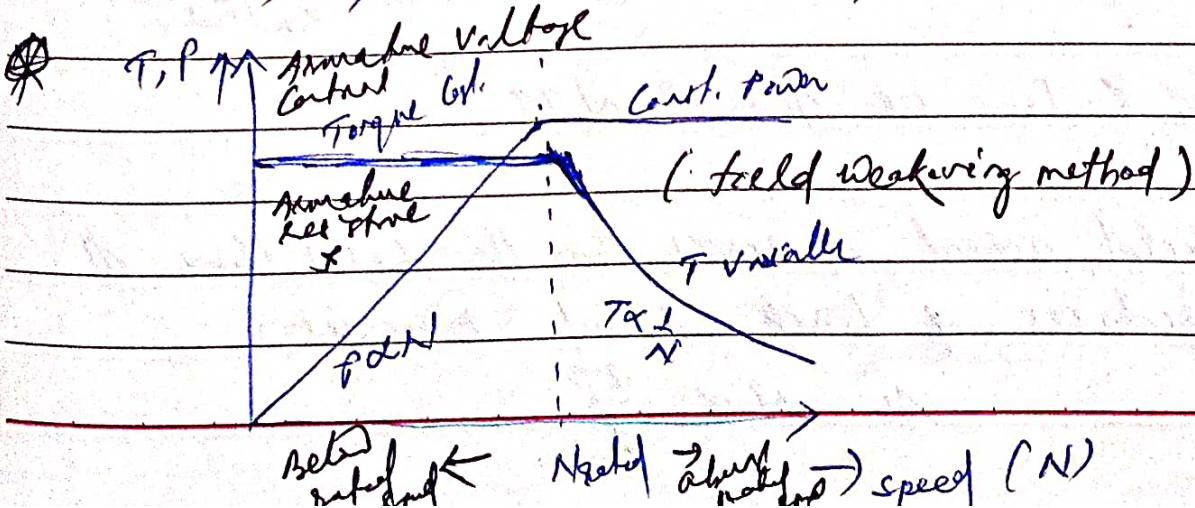
As  $N \propto V$  If  $V$  is increased there is a chance of insulation failure in the winding as insulation of voltage.

→ This method is generally applied to DC separately excited motor. Generally not preferred for DC shunt motor.

Feature :-

i) Constant Torque drive.

ii) speed below base speed.



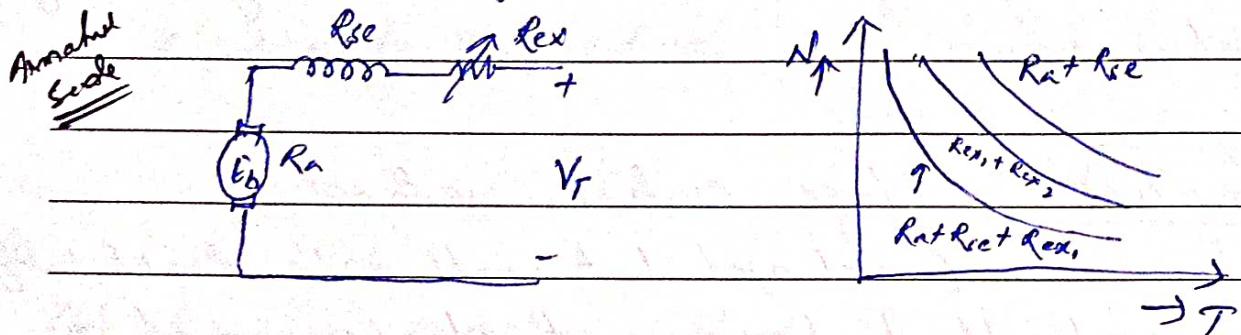


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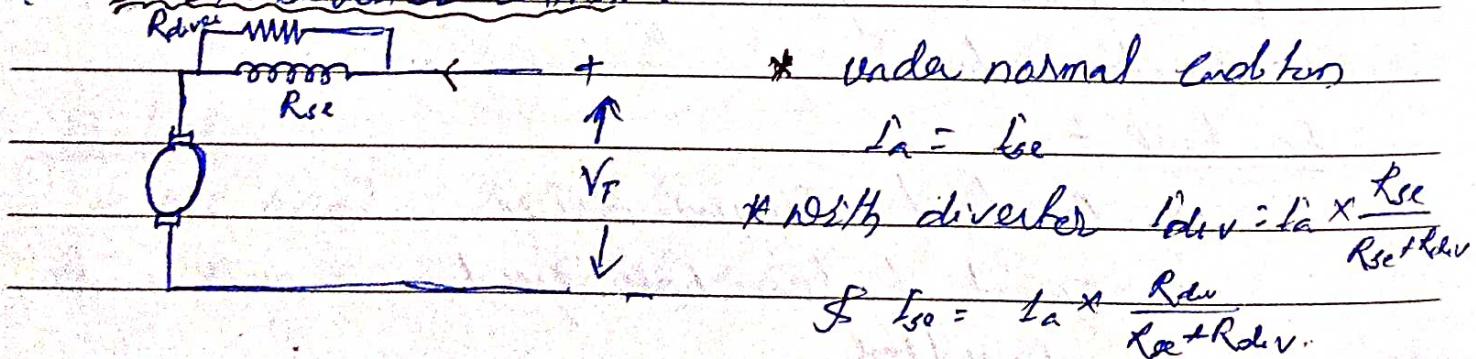
## Speed control of DC series motor

① Armature Resistance Control  $N \propto \frac{1}{R_a + R_{ex}}$



- Field  $\left\{ \begin{array}{l} (2) \text{ Field Diverter control} \\ (3) \text{ Tapped field control} \\ (4) \text{ Series-Parallel control} \end{array} \right.$

(3) Field diverter control :



$$T_{dav} \leq T_a \text{ i.e. } \phi \leq N \tau \tau$$

As field current decreases then the field flux  $\phi$  reduces hence speed increases

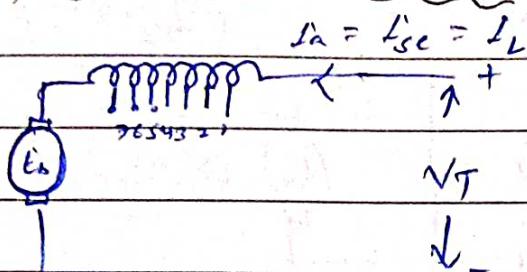
$$\text{As } N \propto \frac{E_b}{\phi}$$



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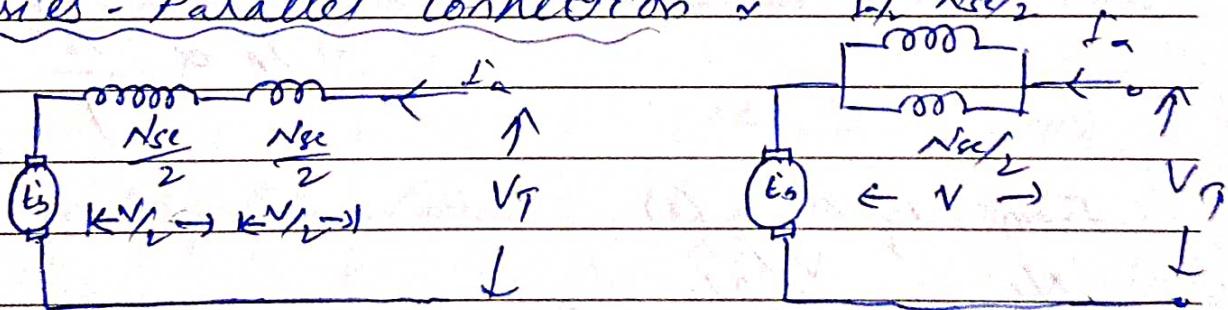
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### (3) Tapped field control.



- \* Tapping is provided in the field winding.
- \* Generally used in electric traction.
- \* When all turns are connected  $\rightarrow$  Lowest speed as flux ~~increases~~ is maximum.
- When turns are cut out  $\rightarrow$  slowly speed will increase.

### Series - Parallel Connection



$$AT = \text{current} \times \text{TURNS}$$

$$\frac{I_a N_{sc}}{2} + \frac{I_a N_{sf}}{2}$$

$$= \frac{I_a N_{sc}}{TTO}$$

$$AT = \frac{I_a}{2} \times \frac{N_{sc}}{2} + \frac{N_{sf} I_a}{2}$$

$$= \frac{I_a N_{sc}}{2}$$

$$\text{ie } |(AT)_{\text{series}}| = 2|(AT)_{\text{parallel}}|$$

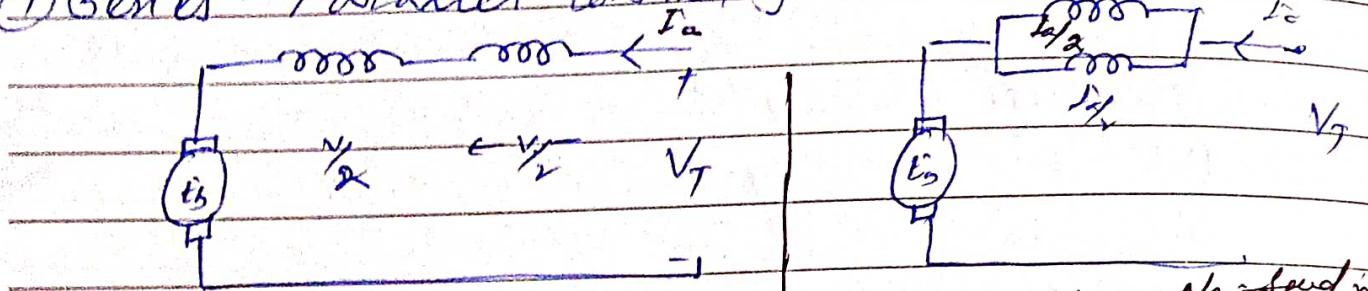
$N_{sc}$  is the  
no. of turns



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#### (4) Series - Parallel Control of DC Series Motor



$$N_S \propto \frac{V}{\phi}$$

$N_S$  = speed in series

$$N_P \propto \frac{V}{\phi}$$

No speed in parallel

$$\Rightarrow N_S \propto \frac{V/2}{I_a} \quad (\text{as } \phi \propto I_a)$$

$$\Rightarrow N_S \propto \frac{V}{2I_a} - \textcircled{1}$$

$$T_d \propto \phi I_a$$

$$\Rightarrow T_d \propto I_a^2 - \textcircled{2}$$

$$N_P \propto \frac{V}{I_a/2}$$

$\phi \propto I_a$   
Equivalent  
current is  $I_a/2$

$$\Rightarrow N_P \propto \frac{2V}{I_a} - \textcircled{3}$$

$$T_p \propto \phi^2 I_a^2 \Rightarrow T_p \propto I_a^2$$

$$\Rightarrow T_p \propto (I_a/2)^2$$

$$\Rightarrow T_p \propto \frac{I_a^2}{4} - \textcircled{4}$$

From Eq. 1 & 3

$$\frac{N_S}{N_P} = \frac{\frac{V}{2I_a}}{\frac{2V}{I_a}} \Rightarrow$$

$$\Rightarrow \frac{N_S}{N_P} = \frac{1}{4} \quad \therefore \boxed{N_P = 4N_S}$$



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From Eq (2) &amp; (4)

$$\frac{T_s}{T_p} = \frac{I_a^2}{I_a^2/4}$$

 $T_s$  = Torque in series connection $T_p$  = Torque in parallel connection.

$$\Rightarrow \frac{T_s}{T_p} = 4 \quad \Rightarrow T_s = 4 T_p$$

### Testing of DC machine -

(a) Losses

(b) Efficiency =  $\frac{\text{Power}}{\text{Power} + \text{Losses}} \times 100$ (c) Voltage Regulation in case of generator =  $\frac{V_{N.L} - V_{F.L}}{V_{F.L}} \times 100$ 

$$\approx \frac{E_g - V}{V} \times 100$$

(d) Speed regulation of Motor =  $\frac{N_o - N_{F.L}}{N_{F.L}} \times 100$ (e) Temperature rise ( $\theta$ ) must be within its permissible limit otherwise may damage insulation

(f) Vibration

(g) Humming / sound etc.



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## Methods of Testing of DC Machine.

- 1) Direct Method - Brake test.
- 2) Indirect method - Swanburne's test.
- 3) Regenerative method - Hopkinson's test.

### ① Brake Test.

- Each machine is designed to operate within certain specified limits.
- The name plate of M/C shows these details as specifications. It indicates the safe operating limits for that machine.
- If these limits are exceeded for considerably long time because of that then this may cause excessively temperature rise and because of this it may cause damage to one or more part of the machine.

\* Rated Power =  $\frac{\text{Max Output power of motor as generator.}}{\text{Specified in SI system of measurement i.e K Watt}}$   
Eg. 3.7 K Watt

\* Duty is operating hours of M/C      continuous duty  
If duty is 15 min       $S_i = \frac{1}{15} \times 24 \text{ hours}$



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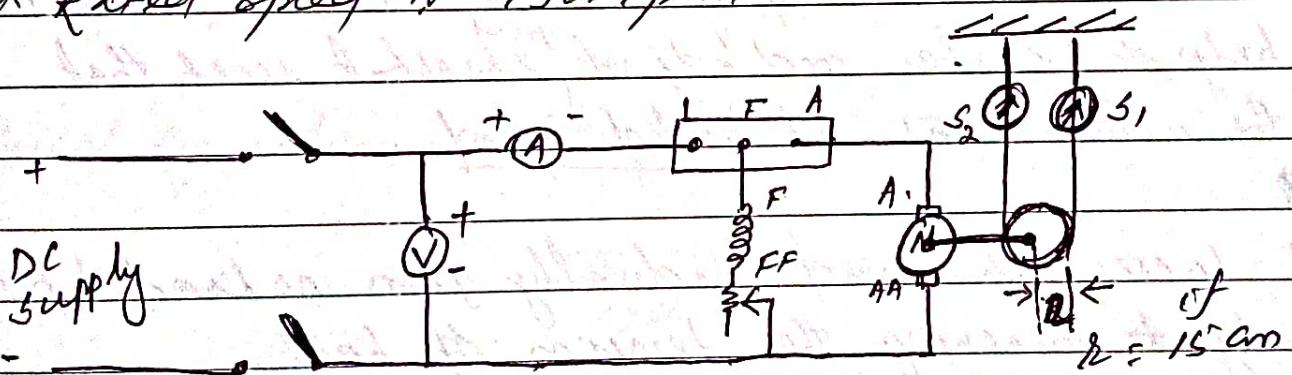
\* Rated armature voltage or Permissible voltage across armature.

\* Rated armature current = Full load current of the m/c Eg. 40A. Generally it can be overloaded up to 120% i.e.  $48 \text{ A}^{40\%}$  (short time period)

\* Rated field <sup>Voltage</sup> ~~current~~  $\approx$  Max. Voltage can be applied across field winding.

\* Rated field current  $\approx$  Eg. 0.87A. max.

\* Rated speed  $\approx$  1500 rpm.



Efficiency can be calculated directly by measuring the input & output power by loading the machine upto full load.

$\Rightarrow$  The machine is loaded by mechanical brake so that the output power is totally wasted.

This method is for small m/c only as all power is wasted.



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- ⇒ A ~~rope~~ belt is wound around the pulley and its two ends are attached to the two spring balances  $S_1$ ,  $S_2$ .
- ⇒ The load on motor can be increased by increasing the tension of the rope.

#### Procedure :-

- \* Note down name plate
  - \* Connect the ckt as shown in ckt diagram.
  - \* Switch on the supply & start the motor with the help of starter and adjust rheostat such that the motor runs at rated speed.
  - \* Increase the load gradually from no load to full load by increasing the tension of belt.
- Record the reading in observation table at each step.

X

#### Observation Table

Sl no	Supply Voltage in Volt	Load current $I_L$ (in Amp)	(3.) $w_1$ (kg)	(4.) $w_2$ (kg)	Speed (rps)
1	230	5	6	2	480
2					
3					
:					



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\* After this gradually reduce the load by decreasing the belt & switch off the supply.

Power Input =  $N \times I_L = 230 \times 5 = 1150$  watt

Shaft Torque ( $T_{sh}$ ) =  $9.81 (S_1 - S_2) b = 9.81 (6-2) \times 0.15$   
= 5.886 Nm.

$$P_{out} = \frac{2\pi N T_{sh}}{60} = \frac{2 \times 1 \times 1480 \times 5.886}{60} = 912.29 \text{ watt.}$$

\* Efficiency ( $\eta$ ) (%)

$$\eta = \frac{P_{out}}{P_{in}} \times 100 = \frac{912.29}{1150} \times 100 = 79.33\%$$

Speed regulation =  $\frac{N_{no} - N_{nL}}{N_{nL}} \times 100$

$$= \frac{1500 - 1480}{1480} \times 100$$

Adv. ~~Adv.~~

- ① All actual loading conditions can be checked.
- ② This gives most accurate results.
- ③ Any type of m/c can be tested.

Dis Adv.

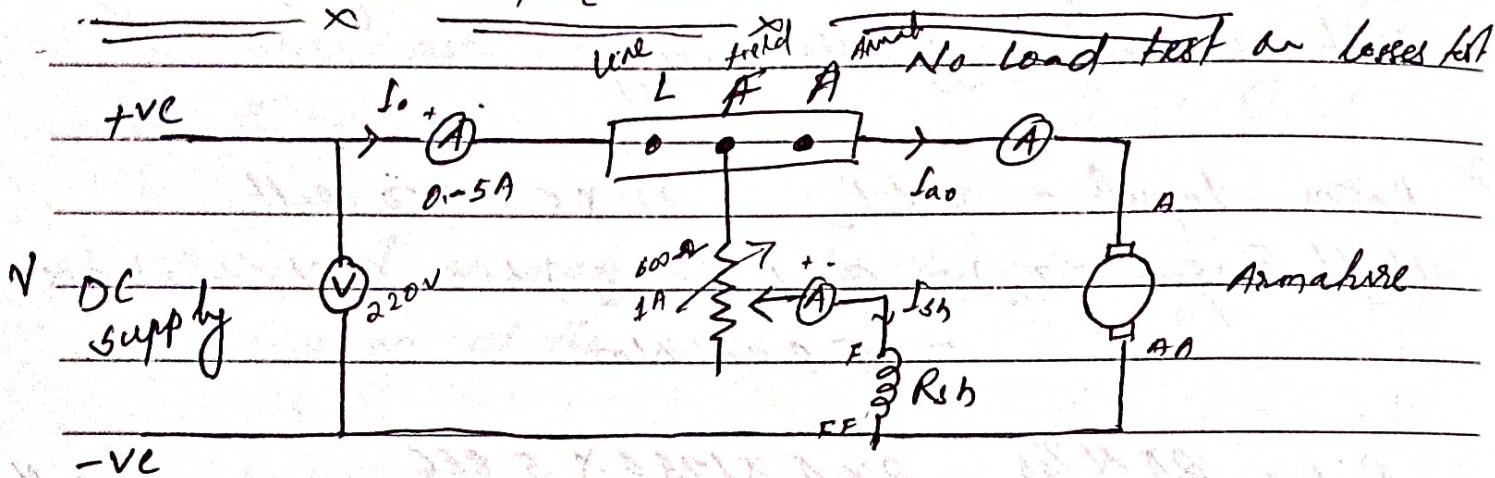
- ① Entire input power is wasted during testing.
- ② Suitable for small dc machine only.



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## Indirect Method (Grainger's Test)



- \* operated under no load condition to determine loss
- \* Power required is very less. (cu loss + lost loss)

Let  $V$  = Supply voltage to the motor.

$I_a$  = No load current

$I_a0$  = No load armature current

$I_{sh}$  = Current of the shunt field.

$R_{sh}$  = Resistance of the shunt field.

$R_a$  = Resistance of the armature winding

$R_{ao}$  = Armature resistance at no-load.

$$\text{At no load} \quad I_{ao} = I_a - I_{sh} \quad \text{--- (1)}$$

$$\& \text{no load power } (P_i)_{\text{no load}} = VI_{ao} \quad \text{--- (2)}$$



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This total power is used to supply certain losses.

But (1) Iron losses in core.

and (2) friction losses in bearing & commutator.  
Also (3) windage losses.

But (4) Armature copper losses at no load.

∴ (5) Loss in field =  $V I_{no}^2 = \frac{V^2}{R_A} R_A$

∴ (6) Stray load loss = Iron loss + friction loss  
Stray loss = + windage loss.

∴ Stray loss = Input power at no load - Armature  
 $P_A$  copper loss at no load - Field copper loss

$$P_s = P_i(\text{no load}) - P_{A0} - P_f$$

Where

$P_s$  = Stray load loss

$P_{A0}$  = No load Armature loss.

$P_f$  = field copper loss.

$$\therefore P_s + P_f = V I_o - P_{A0}$$

But  $P_s + P_f$  = constant loss

Hence we can write

$$\underline{\underline{P_s = V I_o - P_{A0}}} - \textcircled{9}$$



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For motor

Let  $I$  = value of motor current at which efficiency is to be calculated.

$$\text{Motor Input} = VI - (4)$$

$$\therefore \text{Armature Copper losses} = I_a^2 R_a \\ = (I - I_{sh})^2 R_a \quad (\text{for dc shunt motor})$$

$$\text{constant losses } (P_c) = I_{sh}^2 + V_{sh} \text{ or } P_c + P_f$$

$$\text{Hence Total losses of motor} = \text{Armature Cu. loss} + \text{constant loss} \\ = (I - I_{sh})^2 R_a + P_c \quad (5)$$

$$\text{Efficiency } (\eta) = \frac{\text{Output Power}}{\text{Input Power}} \times 100$$

$$\text{or } \frac{\text{Input - losses}}{\text{Input}} \times 100$$

$$= \frac{VI - [(I - I_{sh})^2 R_a + P_c]}{VI} \times 100$$

For General

$$\frac{\text{output}}{\text{output + loss}} \times 100$$

$$= \frac{VI - (I + I_{sh})^2 R_a + P_c}{VI + (I + I_{sh})^2 R_a + P_c} \times 100$$



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\* Adv :-

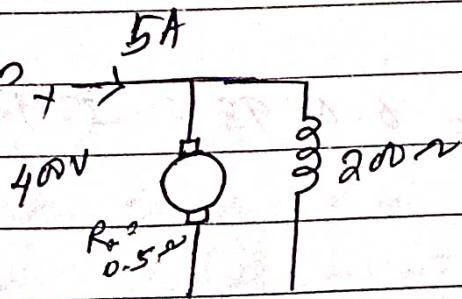
\* Very less amount of power is required to test very large dc machine

\* The test is conducted at no-load hence the permissible temperature rise can't be determined by this method.

\* Only suitable of dc shunt motor not suitable for dc series motor

Q.1 When running as no-load a 400V shunt motor takes 5A, armature resistance is 0.5Ω and field resistance is 200Ω. Find the output of the motor & η when running on full load taking a current of 50A. Also find % change in speed from no load to full load.

All Given



$$I_{fd} = 50A$$

$$\text{Input power} = 400 \times 5 = 2000 \text{瓦}$$

$$I_A = \frac{400}{200} = 2A$$

$$\therefore I_a = 5 - 2 = 3A$$

$$\% \text{ load copper loss} = I_a^2 R_a = 3^2 \times 0.5 = 4.5 \text{ Watt}$$



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$$\text{Total constant losses} = 2000 - 4.5 = 1995.5 \text{ Watt}$$

$$P_c = 1995.5 \text{ Watt}$$

$$\text{Again at full load } I = 50A \quad i_f = 2A \quad (\text{last})$$

$$I_a = 50 - 2 = 48A$$

$$\text{Input} = Vi = 400 \times 50 = 20,000 \text{ Watt}$$

$$\text{Armature losses} = I_a^2 R_a = 48^2 \times 0.5 = 1152$$

$$\eta = \frac{\text{output}}{\text{input}} = \frac{\text{input} - \text{loss}}{400 \times 50} = \frac{20000 - 1152 - 1995.5}{20000}$$

$$= 89.26\%$$

$$\text{for open circuit } \frac{N_1}{N_2} \rightarrow \frac{E_b}{E_b} = \frac{400 - (R_a \times I_a)}{400 - 48 \times 0.5}$$

$$\Rightarrow \frac{N_1}{N_2} = \frac{400 - 3 \times 0.5}{400 - 48 \times 0.5} = \frac{398.5}{376}$$

$$\therefore \frac{N_1 - N_2}{N_2} = \frac{22.5}{376} = 0.0598 = 5.98\%$$



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## Losses & Efficiency

Losses in case of motor is same to that of generator.  
Basically there are 3 types of losses.

(a) Copper losses

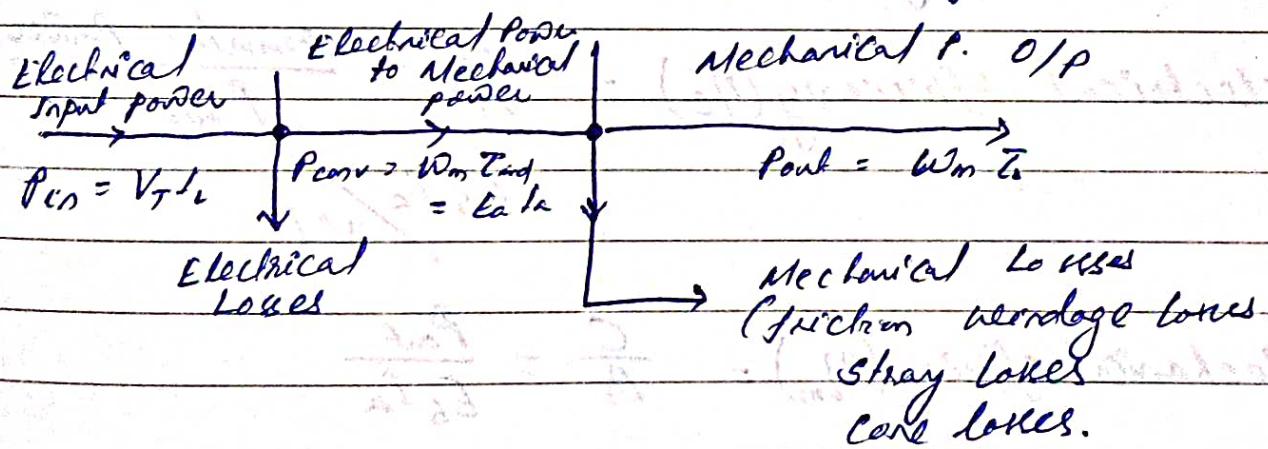
(b) Magnetic losses / Iron losses

(c) Mechanical losses (a) Brush losses

core friction → windage loss } (b) Stray load losses.

The condition for maximum power developed by the motor is  $I_a R_a = \frac{V}{2}$

So the condition for max. efficiency is that the



Condition for maximum efficiency is that the armature copper loss is equal to constant losses ( $W_c$ )

$$\text{i.e. } I_a^2 R_a = W_c$$

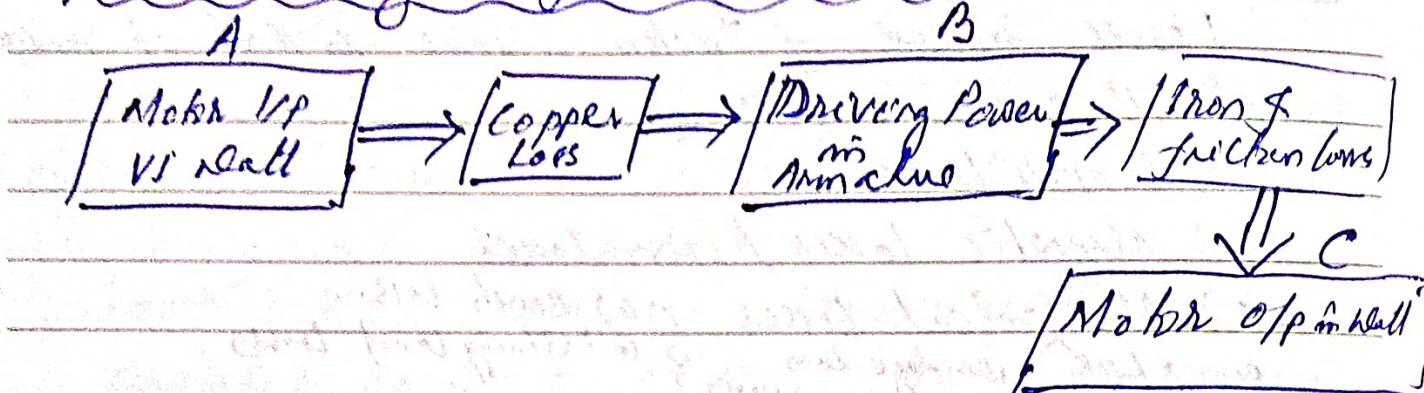
$$I_a = \sqrt{\frac{W_c}{R_a}}$$



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## Power Stages of DC Motor



\* Overall efficiency / commercial Efficiency ( $\eta_c$ ) =  $\frac{C}{A} = \frac{P_{out}}{P_{app}}$

\* Electrical Efficiency ( $\eta_e$ ) =  $\frac{B}{A} = \frac{\text{Armature Power}}{P_{app}}$   
 $= E_b f_a / V I$

\* Mechanical Efficiency ( $\eta_m$ ) =  $\frac{C}{B} = \frac{P_{out}}{E_b f_a}$

## Use of D.C. motor

Dc motors are used in both high & low power drive, for fixed and variable speed electric drive etc.

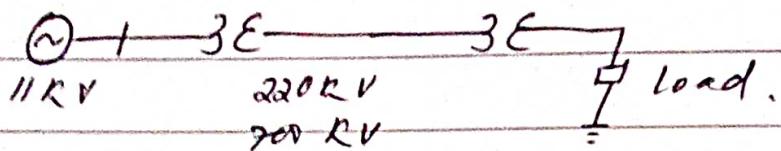
- (i) Electric traction.
- (ii) Electric pumping
- (iii) Cranes
- (iv) Loff
- (v) Air compressor
- (vi) Elevators.



# Instrument Transformer

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Direct measuring method is not possible as these are too high in value.

\* In power systems current & voltage handled are very high and hence direct measurement with conventional instrument is not possible without compromising the operator safety.

→ Solution for this is to stepdown these current & voltages with the help of instrument transformer so that instrument of moderate rating can be used for measurement.

- { The transformer used for measurement of current is called Current Transformer
- { The T/F used for measurement of voltage is called Potential T/F

Instrument Transformer : The transformer used in conjunction with measurement instrument for the measurement purpose is called instrument



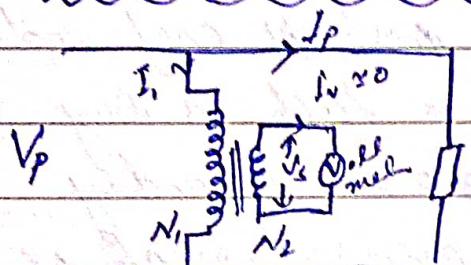
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- ∴ The instrument transformer are used to
- (i) Extend the range of measuring instrument
- (ii) Isolate the measuring instrument from a high voltage line

### operation

#### Potential Transformer (PT)



Similar to power T/F but rating is very less.

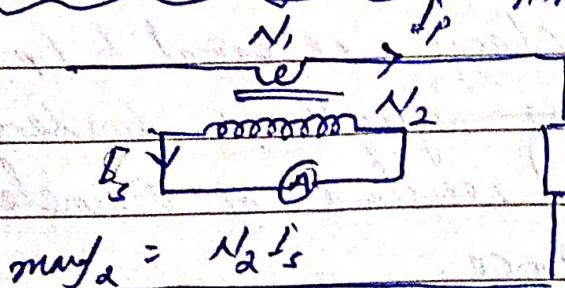
$I_{s \times 0}$  so the power required is very less so the size of T/F required will reduced drastically

∴ The T/F will transfer  $V_p$  to a lesser value in secondary side

$$\text{i.e. } \frac{N_1}{N_2} = \frac{V_p}{V_s} \quad \text{or} \quad \boxed{V_s = V_p \left( \frac{N_2}{N_1} \right)}$$

\* Primary Voltage is fixed.

#### Current Transformer (CT)



\* CT should be such that it must not be super imposed on any extra burden.

\* So the ckt. characteristics is not changed.



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- \* Let  $N_1$  &  $N_2$  be the <sup>number of</sup> turns in primary & secondary side of current transformer.
- \*  $I_p$  &  $I_s$  be current in primary & secondary of CT.

$$\text{mmf produced in primary side mmf}_1 = N_1 I_p$$

So according to Lenz's law mmf ~~or~~ emf induced in secondary side should be in such a way that it will <sup>produce</sup> ~~oppose~~ the mmf ~~or~~ in secondary which opposes the ~~secondary~~ primary mmf i.e mmf<sub>2</sub>

$$\text{i.e } \text{mmf}_1 = \text{mmf}_2$$

$$\Rightarrow N_1 I_p = N_2 I_s$$

$$\Rightarrow I_s = I_p \left( \frac{N_1}{N_2} \right)$$

\* T/F work on the principle of mmf balance ~~so~~ so current in secondary is not infinite even it is short-circuited.

\* Primary current can't change i.e equal to  $I_p$ . Secondary side is having negligible a burden is nearly equal to zero as resistance of Ammeter is negligible or less resistance.



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- The voltage and current of PT & CT standardised at 110V & 5A therefore it is safe for operator & equipment in the switchyard or metering room.
- There are very less power loss in CT & PT compared to shunt and series for extending the range of instrument.
- \* shunt for increasing the current limit.  
series for .. . . . . voltmeter limit.
- Several instrument can be operated from single instrument transformer.

### Ratios of Instrument Transformer

$$\text{Transformation Ratio (R)} = \frac{\text{Primary phasor}}{\text{Secondary phasor}}$$

$$R = \frac{\text{Primary winding current}}{\text{Secondary winding current}} \leftarrow \text{CT}$$

$$R = \frac{\text{Primary secondary Voltage}}{\text{Secondary secondary Voltage}} \leftarrow \text{PT}$$



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Date / /

No. of turns of secondary winding

$$\text{Nominal Ratio } (n) = \frac{\text{No. of turns of secondary winding}}{\text{No. of turns of primary winding}}$$

$$n = \frac{\text{No. of turns of primary winding}}{\text{No. of turns of secondary winding}} \leftarrow PT$$

NOTE

The ratio mentioned on name plate of instrument transformer is nominal ratio.

\* Transformation Ratio (R) = Nominal Ratio (n)  $\times$  Ratio correction factor (RCF)

$$RCF = \frac{R}{n}$$